

FEBRUARY, 1929

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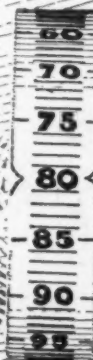
3 SLIP

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IN THIS ISSUE

RADIO MARKETING FOR THE FARMER
EXPERIMENTAL A.C. SCREEN-GRID RECEIVER
AMPLIFICATION OF PHOTOELECTRIC EFFECTS
EFFECTS OF ROOM FURNISHINGS IN SPEAKER QUALITY
SCREEN-GRID R.F. AMPLIFICATION

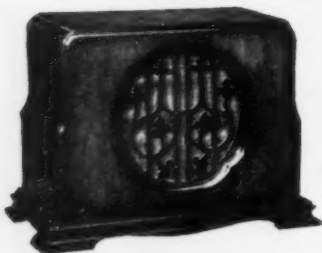


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For AC Operation . . . \$70



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Jamaica Plain, Boston, Mass., U. S. A.



Faradon

Electrostatic Condensers for All Purposes

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1

RADIO

Established 1917

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FORECAST FOR MARCH ISSUE

The principles of the electro-static loudspeaker are explained by "Yose." Frank C. Jones discusses resonance in the dynamic speaker cabinet. John P. Arnold traces developments in synchronization of visual communication and also presents many other interesting facts about radio pictures. R. Raven-Hart interprets foreign circuit diagrams and describes several novel foreign tubes. C. W. Guyatt tells how to make a condenser multiplier for an a.c. voltmeter. Harry R. Lubcke analyzes the design of a practical vacuum tube voltmeter. Carl Joseph describes the construction of a direct-reading ohmmeter. A. Binneweg, Jr., illustrates the construction of a 1½-meter oscillator. Glenn Browning presents some interesting facts about the a.c. screen-grid tube.

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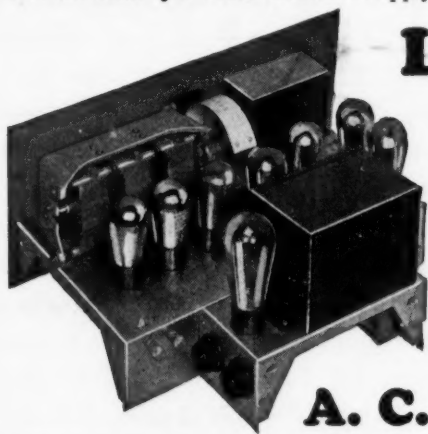
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..You're there with a Crosley..

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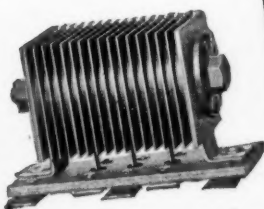
NOW! An Elkon Dry Rectifier for PHILCO POWER UNITS



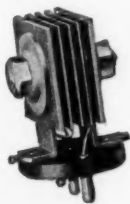
Type BNK for replacing the acid jars in Balkite Types N and K Trickle chargers.



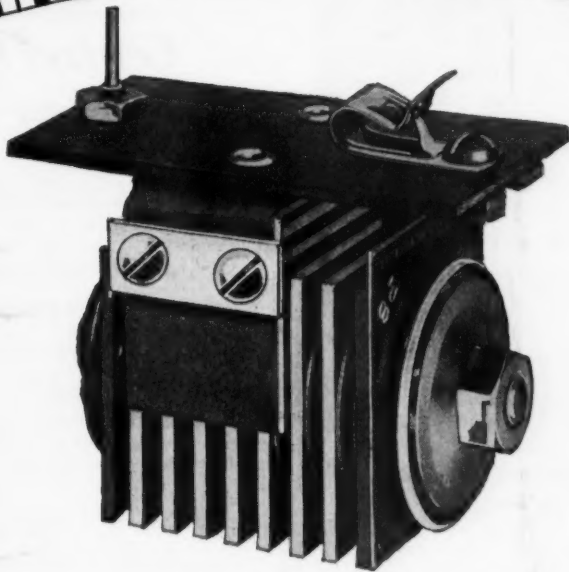
Type RJ for replacing the acid jars in Balkite Type J chargers



Type M-16 for replacing the rectifiers in 11 makes of "A" Eliminators and 3 Ampere chargers



Type V-4 for replacing the rectifiers in 6 makes of Trickle chargers.



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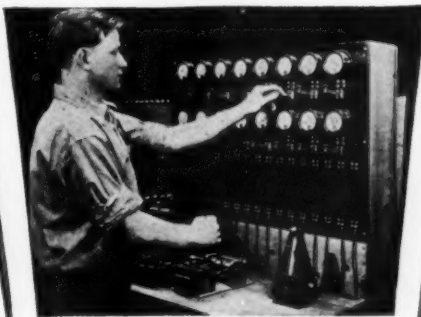
Do not be surprised at its small size—for in its sturdy couples are packed thousands of hours of perfect service. Its compactness is one of the reasons for its efficiency.

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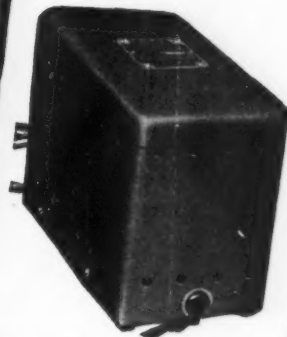
Division of P. R. Mallory & Co.
350 Madison Ave., N. Y. C.



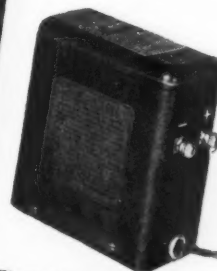
Not a music lesson but one of the many accurate processes of testing Elkon rectifiers.



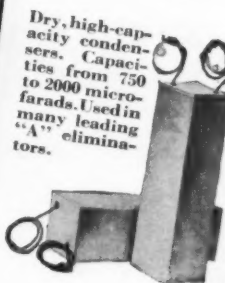
5000 hour type EBH rectifier for replacing BH type tubes in "B" eliminators. Replaces fragile 1000 hour tubes.



3 Ampere Charger. Dry, noiseless, no moving parts. Has tapering feature—long life. For radio or auto batteries.



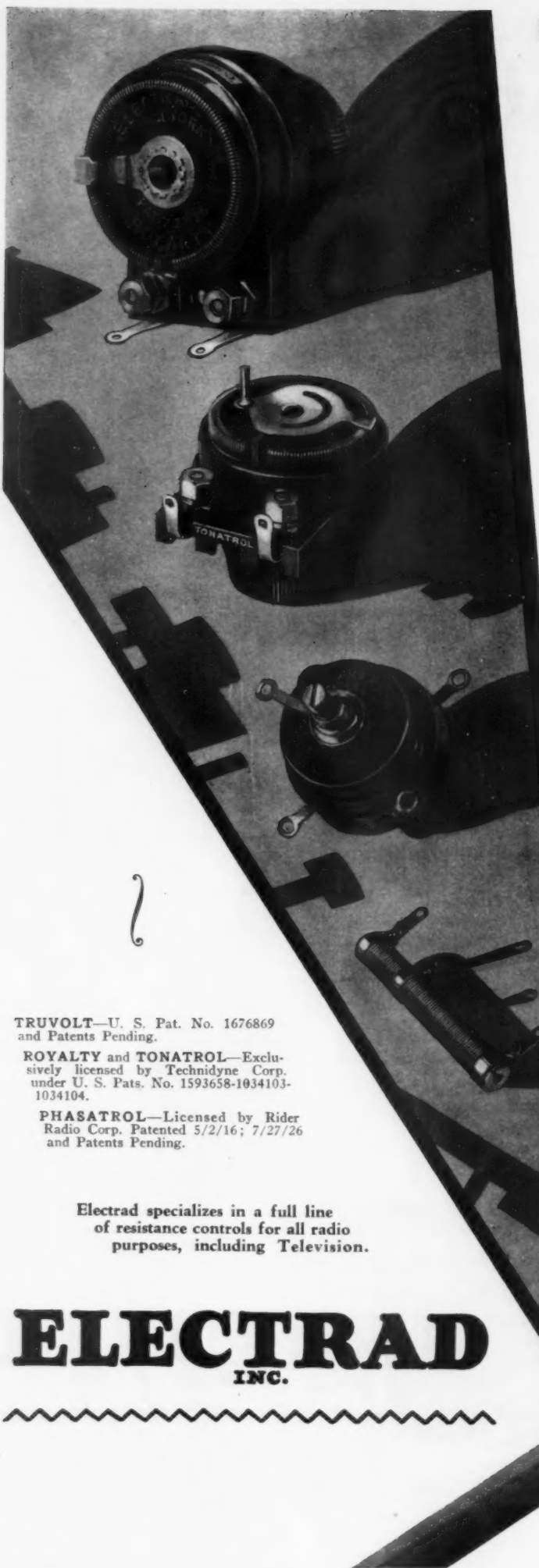
Type E Tapering Charger. Maximum charging rate 1 amp. Dry. No moving parts. Long Life. Ideal for dynamic speakers.



Dry, high-capacity condensers. Capacities from 750 to 2000 microfarads. Used in many leading "A" eliminators.

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TONATROL PHASATROL

Name

Street

City State

RADIO

VOLUME XI

FEBRUARY, 1929

No. 2

Radiatorial Comment

THE popularity of the syndicated comic strip in the newspapers may be rivaled by that of the phonograph record which is supplied exclusively to radio broadcast stations. While these have been used mainly for voice and music, they are also available for rather a crude form of radio movies.

* * *

JEWELERS report a recent increase in trouble from magnetized watches. In many cases this can be traced to the strong magnetic field of an electro-dynamic speaker. This trouble may eventually require the use of a magnetic screen around the speaker. Of course a watch can readily be demagnetized by placing it inside a coil through which an alternating current is passing, but it is liable to be re-magnetized when again exposed.

* * *

THE FEDERAL RADIO COMMISSION has published a long list of signed comments from radio listeners who praise the new allocations. But a possibly equally long list of complaints is not published. Meanwhile the Radio Manufacturers' Association has queried its members regarding their opinions of the new set-up and how it may be improved. This information will be presented to the Commission in the form of a complete survey of national and local conditions.

* * *

IN TELEPHONY, the telephone transmitter is an electrical ear and the receiver an electrical mouth. In television, the transmitter is part of an electrical eye, or light-sensitive device, which converts variation of light into variation in electric current and whose optic nerve is a wire or radio channel which terminates in a device for converting variations in current into variations

in light. The usual converter of light to electricity is a photoelectric cell. The usual converter of electricity to light is a neon tube.

* * *

SHORT waves have resurrected the "good old days" for the broadcast listener. Once more can he strain his ears in order to hear the call letters of distant stations. Again it becomes possible for him to listen to the amateur QRM and to twirl his dials through great empty spaces.

Yet the short waves offer a fascination and interest, even to the BCL, which is lacking in the facile tuning of local broadcast stations. He will enjoy greater freedom from static and will often hear distant stations during the daylight hours. And if he possesses the patience necessary to tune his set and learn the code he will open up new pleasures in radioland.

* * *

DEVELOPMENT of the talking movie was dependent upon radio, because it was not until radio stimulated the demand were sensitive microphones and efficient sound amplifiers developed. With ordinary telephone and sound-recording equipment, the speaker must stand so close to the recorder that no acting can be done. On the other hand, a sensitive microphone and vacuum tube audio amplifier may be placed at such a distance that it does not appear in the picture and the voice and action can be recorded simultaneously.

* * *

SO SUCCESSFUL have been the results of the persistent and energetic campaign of the Pacific Radio Trade Association in reducing interference to radio reception in the San Francisco Bay region that other cities throughout the country are planning similar work. Nearly two

thousand complaints have been investigated and, in a majority of cases, have been remedied during the past year. The work has been financed by the public service corporations whose equipment is sometimes responsible for the interference. Thus neither the radio dealers nor the public has been put to any expense for eliminating these troubles.

The effect has been highly beneficial to the dealers, who are now selling sets where they could not be sold before, and to the listening public who can now enjoy radio without the unseemly noises due to interference. This will also redound to the benefit of the radio industry which is not selfishly allowing the public to suffer after buying radio sets.

* * *

RADIO and aviation typify man's success in piercing four of the seven veils. Nor is this statement purely allegorical. Scientists are gradually adopting the theory that evolution has been coincident with the development of space consciousness. Three of the seven veils that hide the "why" of things are the three dimensions of space: length, breadth and thickness. The fourth is time.

The lowest forms of plant life can be conscious only of the succession of events associated with the passing of time; rooted to one spot they have little or no consciousness of space. The lowest forms of animal life can be imagined as gaining an additional consciousness of space as they move from place to place. The higher forms of animal life are evidently able to think and function in two-space dimensions.

While man has long been able to think in three dimensions, aviation has been his first real opportunity to move in three dimensions. Aviators in the World War have told of watching armies function in two dimensions. While the aviator, from his superior position in the sky, could anticipate inevitable happenings before they occurred, the armies fought blindly because they were unable to see through intervening obstacles.

Radio, by communicating man's thought with the speed of light, has practically nullified both space and time. So now he is ready to conceive of the combination of space and time as the four-dimensional manifold in which world events occur.

AS RADIO develops from an art to a science standards of performance become essential. In the early days a receiver's selectivity was either "broad" or "sharp," its sensitivity was gauged upon its ability to give "coast to coast" reception, and its audio quality was either good or poor.

While these loose terms sufficed for the early state of the art, they are no longer acceptable. How broad is the tuning? What is the overall voltage amplification? What is the gain in deci-bells or transmission units for each of the audio frequencies?

These are questions which can now be definitely answered on the basis of precise measurements. This has recently been made possible by the development of new methods and more delicate apparatus than was required for the older electrical measurements. Many radio measurements in receiving sets are of currents and voltages which are less than one-millionth those used in electric light and power practice. A radio receiver is virtually a power converter whose output in milli-watts is proportional to its input in milli-volts.

The methods used for making these measurements have been described from time to time in the *Proceedings* of the Institute of Radio Engineers. So there is no reason why any radio manufacturer cannot specify that his receiver has a selectivity of such-and-such per cent ten kilocycles from resonance, and a sensitivity such that a stated minimum voltage input is converted into a standard power output, or a gain of so many D. B.'s at 60, 600, and 6000 cycles.

But as yet no authoritative organization has set up what may be accepted as standards of performance. Here is a task that may well be undertaken through the coöperative effort of I. R. E., R. M. A. and Nema committees. Thus it will be possible for the well-informed purchaser to know the relative performance of two receivers whose characteristics are technically defined.

Such standardization would tend to eliminate many careless statements in advertising and develop a public confidence which is now lacking in many of the exaggerated claims that are made by salesmen. Consequently the radio industry would acquire a greater stability. 'Tis a summation devoutly to be wished!

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Radio Marketing for the Farmer

A CENTRAL CALIFORNIA farmer was recently offered 60 cents a bag for his crop of onions by an agent who seemed very eager to buy. Suspecting that the buyer was armed with a superior knowledge of the marketing situation, the farmer got in touch with the Federal-State Marketing Service radio station, KRB, at Salinas, California, asking for a report on crops and current prices. KRB referred the question to the San Francisco station, KRG, where trained market analysts investigated the onion situation and reported at once that onion crops had been badly damaged in other sections and that the demand was on an upward trend. KRG sent the report back to Salinas and the farmer, no longer at a disadvantage, refused the offer of 60 cents, later selling at \$1.65.

This is a typical example of what the U. S. Department of Agriculture and the California Department of Agriculture are doing for the California farmer. Guided by Mr. B. H. Critchfield, Chief of the Division of Markets, the Federal-State Marketing News Service gathers, analyzes and disseminates information regarding the eastern and foreign markets; the supply and demand of farm produce. Effectually, they are equalizing the bargaining power of the farmer with that of the produce buyer.

Six thousand men are in the field to gather this information and send it to Washington for analysis. From here it is sent to San Francisco by leased wire, whence it is relayed to all points in the short-wave radio network.

KRG, the control station at San Francisco, is a 500-watt station, operated on a wavelength of 32.4 meters. The outstanding features of the equipment are its commercial appearance and workmanship combined with its adherence to the low loss principle.

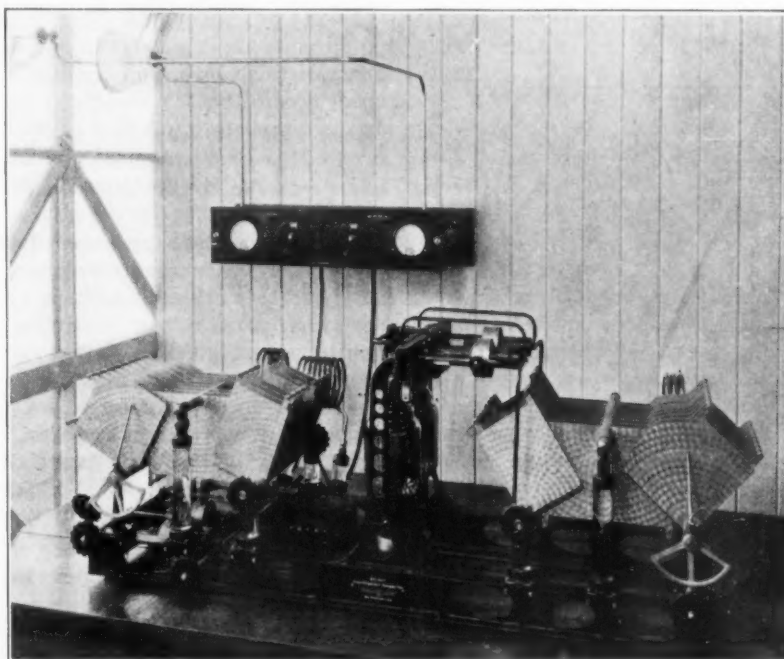
The transmitter is mounted on a cast-iron frame, the separate circuits being isolated from each other with specially

moulded pyrex insulation. The condensers and coupling controls are operated by step-down gears, the scales on the rotor plates passing in back of a marker. The two meters on the antenna condenser panel are thermo-couple ammeters in the antenna and counterpoise circuits.

On the power control panel is mounted the filament and plate switches, filament

resulting oscillation, modulated at about 1000 cycles, is very steady and pleasing to copy.

The transmitters at the field stations are 100 watts in power, self-rectified, and are more or less portable so that they may be moved during the various seasons to more active agricultural centers. A great deal of experimental work was carried on by the engineer in charge, Mr.

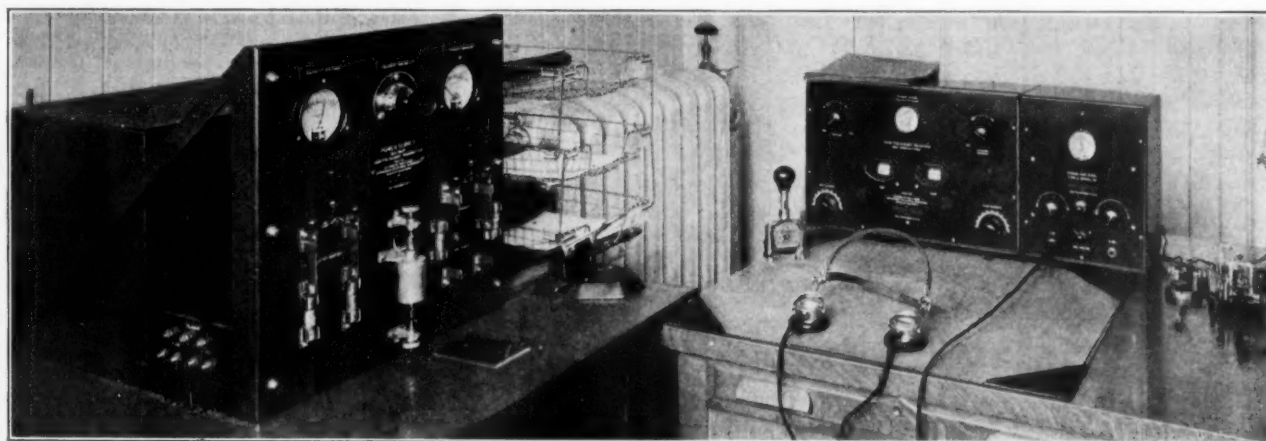


500-Watt, 32.4 Meter Transmitter at Control Station KRG

voltmeter, input ammeter and high-speed keying relay, which breaks the circuit in the primary of the plate transformer. This is a plunger type relay, designed by Ralph Heintz, builder of the apparatus, for continuous operation with a minimum of heating. A tuned-plate tuned-grid, self-rectified circuit is used, employing two UV 204-A tubes which are supplied with 500 cycle a.c. from a 2 k.w. General Electric alternator. The

Herbert J. Breuer, prior to applying for a certain frequency, in order to determine which was best suited for daylight communication over specified distances. The final choice of 32.4 meters is found to carry much better to Los Angeles and El Centro than to shorter distances such as Sacramento. This is due to the skip distance phenomenon characteristic of short-wave transmission.

The receivers at all stations are tuned-



Power Panel and Receiver at KRG

grid tuned-plate detectors with two stages of a.f. amplification. With plug-in coils they will cover the range of from 10 to 100 meters.

The operators holding down the various stations are on the job from 7 a. m. until 4 p. m., ready to copy the bulletins



Map of Radio Network

that are constantly pouring into San Francisco from Washington to be transmitted to the network by KRG. The daily traffic over KRG has been averaging 3500 words, most of which is handled at from 30 to 38 words per minute, the operators being among the fastest in the profession. Between bulletins from the eastern and foreign markets the air is filled with reports from the farmers

themselves, weather reports and requests for information.

And so, by the use of short-wave radio the California farmer is being supplied with hourly information regarding the supply and demand, current prices and trend of the markets for his produce in all parts of the world. He is now able to intelligently distribute his crops, and retain for himself and his community the largest part of the consumer's dollar.

PICK-UP ADJUSTMENT

By THOMAS F. McDONOUGH

IGNORANT use of phonograph pick-up units has caused much unnecessary damage to phonograph records. Too much needle pressure wears out the record. Improper lining-up undercuts the record grooves. Blunt needles impair the tone value and ultimately ruin the record. What is the correct usage?

The correct needle pressure is between 5 and 7 oz., depending somewhat upon the make of the record and the type of recording. This is best controlled by a counterweight, whereby the pressure remains nearly constant in spite of any wobble in the turntable. The proper weight may be checked with a small postal scale, the needle resting upon the scale at a level corresponding to the surface of the record.

The counterweight can be conveniently arranged as a slider with a thumbscrew to hold it after it has been moved into the best place. This is the minimum pressure under which the needle maintains contact with the bottom of the groove in the record without jumping. This action can be observed by means of a small microscope and a suitable light.

Records where bass notes predominate usually require more pressure than others. Different makes of records require different pressure. The careful man will label each of his records to indicate the counterbalance setting that gives the best results.

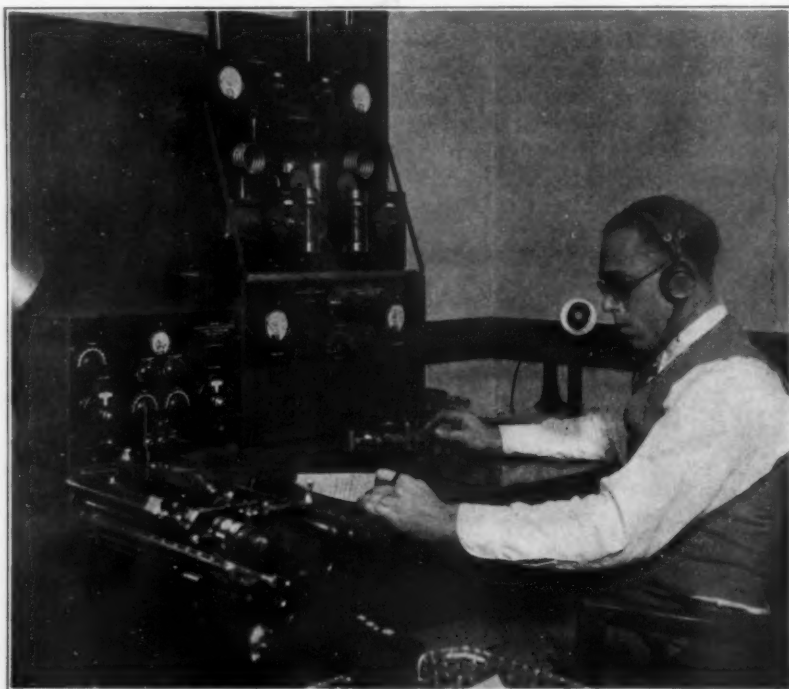
To avoid under-cutting the pick-up ought to be so placed that a line drawn from the point of the needle to the center of the tone-arm pivot is tangent to the record groove at the needle's point of contact in the groove. As there are practical limits to the length of tone-arm which would make this possible, the tone-arm should be placed so that there will be a minimum angle (viewed from above) between the needle axis and any of the record grooves at the point of contact.

The axis of the needle should lie in the vertical plane passing through the line drawn from the needle point to the tone-arm pivot. Furthermore, the needle should not be vertical but should be inclined at an angle of about 65 degrees (viewed from the side) to the face of the record. This angle should be such that the needle will not dig in but will glide smoothly.

A new needle should be used each time a record is played. New needles have smoothly rounded points that fit the bottom of the groove properly. They rush along the groove, brushing and sliding past all the side-bumps and corners which represent music, at a pace of from 150 to 200 feet per minute. The pressure on the bottom of the groove often exceeds 30 pounds to the sq. in., and it requires as much energy to drag the needle past the side bumps as is consumed in normal friction on the bottom of the groove. By the time the needle reaches the end of the groove it has been worn blunt on the end. So blunt in fact that it can no longer follow the bottom of the groove closely and instead of sliding around the side bumps it starts "riding the high spots" so to speak. This means that the upper edges of the side bumps are taking most of the strain. In time they will be worn off until the tone value of the record has been impaired or, if the abuse continues, they will give away entirely and the needle will "take off across lots" when it comes to that point and another record will have "gone west."

Records should always be kept clean. Even finger marks contain a trace of oil to which dust will cling. Wipe the record carefully, both before and after use, with a soft brush. Brush along the grooves, not across them.

When the adjustments have been made as here outlined and the foregoing precautions observed, you may be surprised to find the needle scratch conspicuous by its absence. The quality and fidelity of tone should show a marked improvement also.



100-Watt Field Station KRM

Furnished Rooms

By KEITH LABAR

Illustrated by LOUIS McMANUS

IN THE days when weary clerks assured customers that it was unnecessary to have a switch to shut off the B battery all that was demanded from a radio set was noise. A "powerful set" was a rival to the boy saxophonist next door.

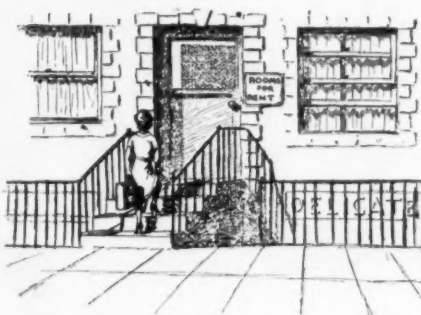
Now we demand tone quality and are even progressing to the point where we do not consider that we have that quality if the announcer sounds as if he had taken up residence in a rain barrel. The improvement in quality has resulted from the improvement of equipment all along the line, from the microphone to the speaker. Music, after being carefully led into the microphone, emerges from the speaker without being much the worse for wear.

At this point everyone is satisfied; the engineers consider the job done and go home. And indeed, what more is there to it? The music is coming out of the speaker—good music. It has lots of umpa, umpa. All those low notes. And there is enough volume.

But we do not hear the music by putting our heads into the set. A pair of phones is rarely used for broadcast reception. We hear music in a room, and upon the design and construction of that room depends the quality of the music that reaches our ears.

Reverberation occurs in large auditoriums, especially, when there is a small audience. Many a theatre has gone on the rocks because of poor acoustics but theatres with heavily carpeted foyers, upholstered chairs, and special plaster decorations have acoustics that manage to get by.

But the same conditions apply to the



small living room of the home as are found in large auditoriums. An architect should submit his plans to an acoustical engineer in order to have the acoustics good for radio or piano. Your daughter's expensive voice will sound better in such a room. If a good engineer had been consulted about the "House of Usher," the regretted occurrence might not have happened there.

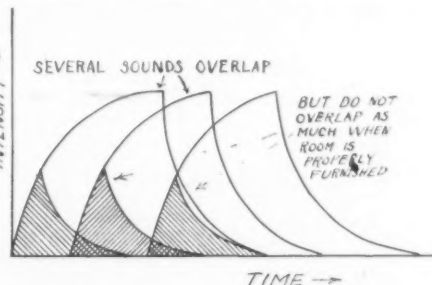
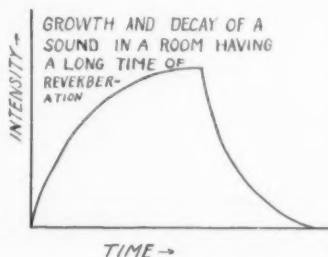
A sound wave travels with a definite speed in air and hits walls or objects which absorb all or part of the sound, depending on the material. Let us say that the sound is nearly all reflected, only a small portion being absorbed—a condition found in buildings with hard plaster walls. It will then go to the

opposite wall, be reflected again and the process will continue until all of the sound is absorbed.

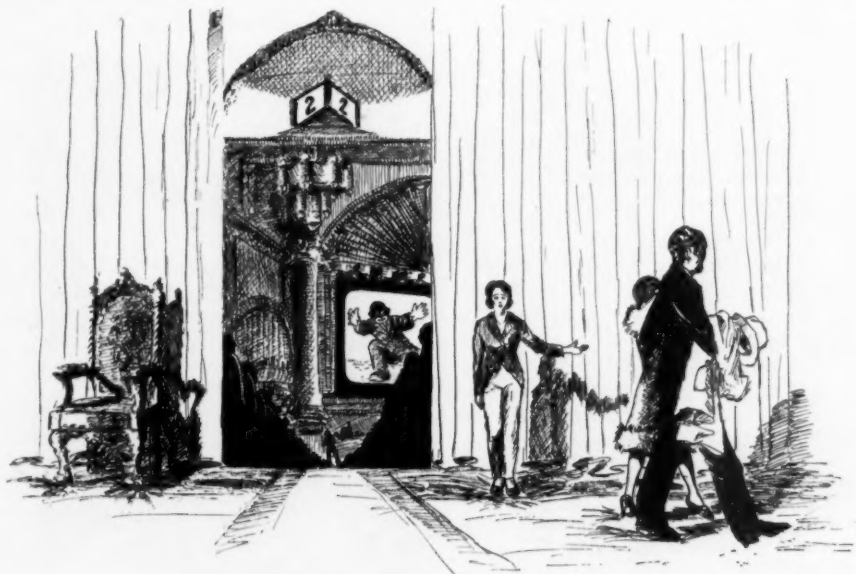
A sound (one wave or more) builds up and dies out in the manner shown in the accompanying diagram. The second wave overlaps the first, due to the first lingering on and dying a slow death. When the series of sounds is music, the effect is not as displeasing as when there is speech. Overlapping sounds of speech are hard to follow. Remember the good old days when the train announcer would come out on a balcony and in a voice that boomed back from every wall announce the train? 'Member? His children are still announcing.

If we soak up much of the reflected sound by hanging curtains or using sound-absorbing plasters, the overlapping is diminished, as is also the intensity. Even though the actual volume of sound that reaches the ear is lessened, the speech is more understandable.

The reader has perhaps realized that there is some relation between the size of the room and the time the sound should be allowed to bounce around unmolested. The time of reverberation should be from 1.00 to 1.05 seconds for



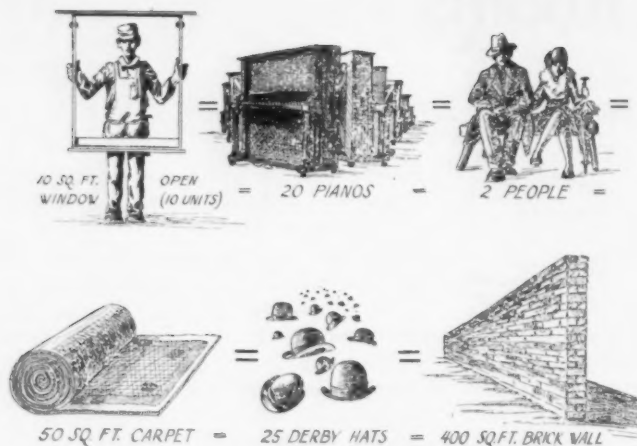
How a Sound Builds Up and Dies Away



Poor Acoustics Have Put Many a Theatre on the Rocks

the small rooms usually found in our homes. The time can usually be found by the formula: $time = .05V \div A$, where V is the cubic feet in the room, and A is a measure of the total absorbing power of the room. If A is small, not much sound being absorbed in walls or otherwise, then the formula shows that the time is large. And again, the larger the cubical size, the longer the time is, for it takes the sound a longer time to get from one wall to the other to be reflected. The factor .05 is merely put in to make the units come out right. It would be different if we measured in fathoms or other cross-word puzzle units of length.

A is expressed in square feet of a perfectly absorbing medium, such as an open window. For a wall, for instance, it is equal to the area of the wall times the absorption coefficient of the material



Typical Sound Coefficients



The Room Whose Sound Coefficients Are Figured in the Text

in the wall, or rather, how many square feet of open window is equal to our wall? Those who have calculated the number of rolls of wall paper required to paper a room can hunt for pencil and paper. The equivalent amount of open window for each surface or article in the room, all added up, gives the total number of units of A to be substituted in the formula.

Suppose you dwelt in marble halls. Marble absorbs 1 per cent of the sound that hits it, therefore coefficient .01, and reflects 99 per cent. Let us say the dimensions are 100 ft. x 50 ft. x 25 ft., a typical Hollywood bathroom. The calculations would then be:

2 walls 100 x 25 5000 sq. ft. x .01 = 50 units
2 walls 50 x 25 2500 sq. ft. x .01 = 25 units
1 ceiling 50 x 100 5000 sq. ft. x .01 = 50 units
1 floor 50 x 100 5000 sq. ft. x .01 = 50 units

Total 175 units

$t = .05 \times 100 \times 50 \times 25 \div 175 = 35$ seconds.

Try to put a radio set in such a room, when a sound would be just audible 35 seconds after the set was shut off.

But let us cover up the marble walls and ceiling with a trick plaster having a coefficient of sound absorption of .20 and the floor with a heavy rug with an



The Transmitter and

absorption of .30, and what do we get?

Total walls and ceiling—
12,500 sq. ft. x .20 = 2500 units
Floor 5000 sq. ft. x .30 = 1500 units

Total 4000 units

$t = .05 \times 125,000 \div 4000 = 1.56$ seconds.

This is a good value for such a large room.

All of these calculations are on the basis of middle "C" on the piano. We assume that the materials take the same helping of low notes as high. Practically, this is not the case. The high notes are absorbed more easily than the low ones. Low notes are absorbed by thick materials, such as a thick rug, overstuffed chairs, curtains hanging in folds and so on. But do not let this worry you.

Now that we are getting into this new game in earnest we need a table of the materials and objects frequently found.

SOUND ABSORPTION COEFFICIENTS

Carpets, unlined	.15 per sq. ft.
Carpets, lined	.20 " " "
Concrete	.015 " " "
Curtains, cloth	.15 " " "
Curtains, velour	.25 " " "
Books in shelves	.20 " " "
Glass	.027 " " "
Heavy rugs	.25 " " "
Linoleum	.03 " " "
Lime plaster	.04 " " "
Oil paintings	.28 " " "
Open window	1.00 " " "
Plaster on tile	.025 " " "
Soft wood	.06 " " "
Varnished wood	.03 " " "

OBJECTS, PER OBJECT

Audience, per person	4.2
Divan	14.0
Piano	.5
Overstuffed chair	8.

The calculations for the room shown in the illustration are as follows: Length 20 ft., width 14 ft., height 9 ft., volume 2520 cu. ft.

ABSORBING POWER

Ceiling 25 x 14, 280 sq. ft. x .03	8.4 units
Rug, 9 x 15, 95 sq. ft. x .20	19.0 "
Bare floor 280-95, 185 sq. ft. x .03	5.5 "
Wall with windows 9 x 14 = 126 sq. ft. x .03	3.1 "

Note: Glass has same "a" as this plaster and so glass is not calculated separately.

The curtains (estimated) 20 sq. ft. x .25 5.0 "

Wall toward us, considered bare 9 x 14, 126 sq. ft. x .03	3.7 "
Wall with archway 20 x 9 = 180 minus archway, area 30, 150 sq. ft. x .03	4.5 "
Archway, 20 sq. ft. x 1.00	20.0 "
Wall where people are sitting 180 x .03	5.2 "
Oil painting 3 x 2, 6 sq. ft. x .28	1.5 "
Divan	15. "
Chair	8. "
Table	1.0 "
Radio	.5 "
Deck	.5 "

Total 100.9 units

This is equal to a square open window ten feet on a side. $\text{Time} = .05 \times 2520 \div 100.9 = 1.25$ seconds. This is a little too much. To cut it down to 1.05 requires that 19 additional units of absorption be added to the room. This can be obtained in the following manner: Three people and a baby, or 76 square feet of velour curtains, or open a window at the end out of the picture, 4 ft. x 5 ft. Or we could use a different plaster.

The small items need not be calculated as carefully as the larger items. The dimensions of the archway are very important, as the effect is large, but we could err in estimating quite a few of

(Continued on Page 32)



the Receiver

A Five-Tube A. C. Screen-Grid Receiver

A Discussion of the "Power" Detector as Used in an Experimental Receiver of Excellent Audio Quality

By FRANK C. JONES

SCREEN-GRID tubes may be used in a set as r.f. amplifiers and as a space-charge detector with only one stage of power audio amplification. A so-called power detector is used to work the power tube to its full limit.

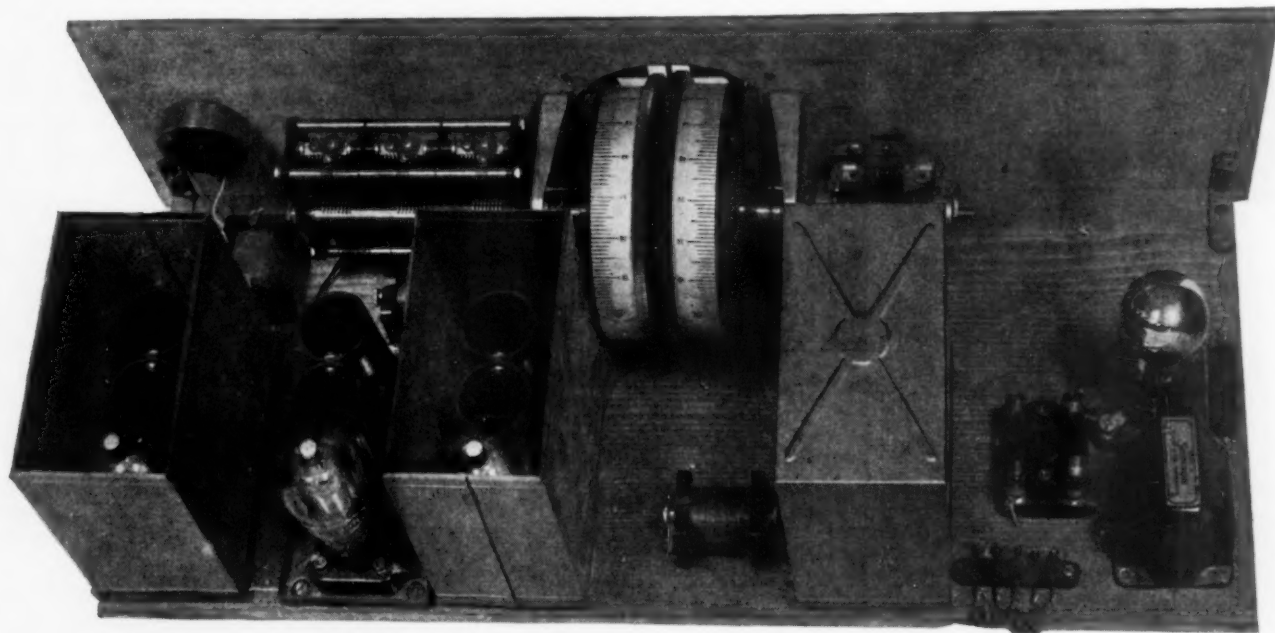
This is accomplished by putting a high grid bias on the detector and by using one more stage of r.f. amplification than would be necessary if an ordinary detector and two-stage audio amplifier were used. All of the high frequency components of the detector output are

use of even five or six tuned circuits which are each rather broadly tuned will give a much better band-pass effect wide enough to pass side-band frequencies as high as 5000 or 6000 cycles per second above or below the carrier frequency.

The screen-grid tube as a space-charge type detector does not attenuate the higher audio frequencies in the process of detection as does a grid leak-condenser detector. For this reason the resulting quality from a good dynamic speaker is quite brilliant and has not the muffled

pass condensers are mounted inside of copper cans. In the picture the sides of one can were left off to show a little of the layout. The cans are made of thin sheet copper and are about 3 x 5 x 5 3/4 in. in size. Standard parts obtainable on the market may be used throughout.

The r.f. coils were of the double solenoid type to minimize the field in order to use the small size of shielding cans. These coils were wound with No. 28 enameled wire spaced in grooves about 45 turns per inch. There were a total



Rear View of Five-Tube A.C. Screen-Grid Receiver

bypassed to ground by a .0001 mfd. condenser in the plate circuit of the detector tube.

Eliminating one audio stage gives better quality with less hum. Even with a.c. filament supply for the '71 tube this receiver gives less hum with a 30-cycle cut-off than is heard from some a.c. receivers with a 200-cycle cut-off. Furthermore any distortion due to the first stage of audio as well as the phase distortion due to use of audio transformers is eliminated in this set.

By using three stages of screen-grid tubes as r.f. amplifiers the r.f. gain necessary for the usual distant reception is obtained. It is generally necessary to have at least four tuned circuits in order to obtain sufficient selectivity. The use of less than four tuned circuits generally means that the tuned circuit losses have to be so low that the upper edges of the side-bands are greatly attenuated. The

sound that many sets have. The low note reproduction is excellent with this detector.

The experimental set shown in the picture was first built for d.c. tubes and then modified for a.c. operation. The set is so satisfactory for a.c. operation that it should be built up that way using a.c. heater type screen-grid tubes. There is some possibility of a.c. modulation if the heavy filament type of a.c. screen-grid tube is used. Screen-grid tubes are recommended due to the gain of 15 to 25 per stage which is one and one-half to two times per stage that obtainable with a multistage amplifier using three-electrode tubes. Three stages of the screen-grid tubes give as much or more gain than a good four-stage r.f. amplifier using ordinary tubes.

The picture shows fairly well the layout of the parts. The four r.f. transformers and screen-grid tubes with by-

of 130 turns, 65 on each tube, which were clamped nearly together. The tubes were 1 1/2 in. in diameter and 2 3/4 in. long. The plate taps should be about 35 turns from the filament end. Tapping higher tends to give more gain, less selectivity, and is also liable to cause oscillation. If a standard solenoid coil is used, it should be tapped at about 3/10 of the way up from the filament end.

Using more turns does not increase the gain per stage very greatly unless a tuned impedance is used, since there is a quadrature current which is not exactly in phase with the induced current, especially at the lower end of the broadcast band.

Using an autotransformer connection seems to give the greatest coupling of primary to secondary for the same degree of selectivity. Greater coupling means greater amplification. The problem of

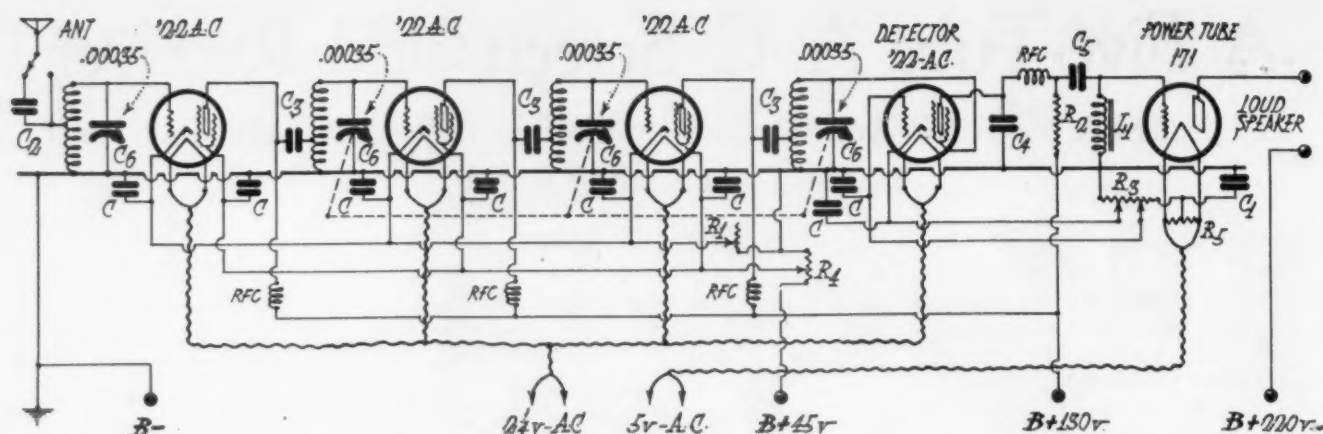


Fig. 1. Circuit Diagram of Five-Tube A.C. Screen-Grid Receiver

R_1 = 0-1000 ohm adjustable resistor.
 R_2 = $\frac{1}{4}$ megohm leak.
 R_3 = 2000-ohm strip with two adjustable taps.

R_4 = 10,000-ohm variable resistance (volume control).
 R_5 = 30-ohm center-tap resistor.

L_1 = 400 henries.
 C = .1 mfd. bypass condenser.
 C_1 = 1. mfd. bypass condenser.

C_2 = .0001 mfd. bypass condenser.
 C_3 = .006 mfd. bypass condenser.

C_4 = .0001 mfd. bypass condenser.
 C_5 = .04 mfd. or larger.
 C_6 = .00035 mfd. variable condenser.

whether a separate primary or an auto-former connection over the whole broadcast band is open for much discussion and experimental work. The number of tuned circuits, losses in each circuit, the amount of shielding, the desired selectivity and gain, and physical dimensions of the coils are all factors which should be considered in the design of any receiver.

Standard .00035 mfd. single and three-gang variable condensers were used. There are two drum tuning controls and a volume control. A 110-volt switch should be located on the panel for convenience. An old *B* eliminator was used for plate supply since it gave about 220 volts on the highest tap and, because of the small drain the adjustable 90-volt tap gave 130 volts for the screen grids. The filaments were heated by means of a small filament type transformer. Since the set was only an experimental set, the 110-volt switch was not mounted on the panel.

The panel was 7 x 24 x $\frac{1}{4}$ in. of three-ply walnut veneer. The baseboard was 23 $\frac{1}{2}$ x 9 x $\frac{1}{2}$ in. of dry spruce. Undoubtedly the set could have been mounted on an 18-in. panel without disturbing the electrical properties of the circuit.

The r.f. chokes should have an inductance of at least 80 millihenrys in order that their effect will be negligible when placed across the primary winding of the r.f. transformers. A similar choke should be used in the plate circuit of the detector, since there is a very strong r.f. voltage which would tend to overload the power tube.

The screen-grid detector requires from 10 to 15 volts bias on the input grid for bias detection at the other voltages used. By using the arrangement shown in the wiring diagram for obtaining bias voltages for the '71 tube and detector, the

proper adjustments for maximum sensitivity and quality are obtained. R_3 consists of a 2000 ohm resistor with two adjustable sliders for the cathode and space charge grid connections. The latter should be from 10 to 20 volts positive with respect to the cathode and the cathode about 12 volts positive with respect to *B*—. The drop across the whole resistor gives the proper bias for the '71 tube. Incidentally the '71 tube should not be used when its emission begins to drop. If the '71 tube does not draw somewhere near its rated current, the bias voltages on the detector tube will not be correct.

The audio coupling unit consists of a 250,000-ohm resistor R_2 for connecting the plate voltage to the detector, and a resonant circuit coupling to the grid of the power tube. This circuit resonates at about 40 or 50 cycles per second when using a blocking condenser of .04 mfd. and a grid audio choke of about 400 henries. A grid choke is somewhat preferable to a grid leak as the '71 tube does not block when overloaded occasionally. Using a choke means a slight drop at the lower audio frequencies because of its lower impedance for lower frequencies. By using the principle of resonance there is an increase of voltage across the condenser and choke due to the increased current. The sharpness of resonance depends upon the series resistance of which the R_2 of the tube and R_2 are in parallel. This resistance is so great that the resonant action is not very noticeable though it does bring up the low notes a little. It is difficult to test by ear because few broadcast stations transmit notes low enough to check this action.

Bias for the r.f. amplifiers may be obtained by means of the 1000-ohm semi-variable resistance in the common plate return to the cathodes, or center tap of the filaments if the heavy filament type of tubes are used. Volume control is obtained by varying the positive bias on

the screen-grids from 0 to 45 volts by means of a 10,000-ohm potentiometer. This arrangement controls the gain of all three stages so the tube noises are minimized for local reception. It also gives very positive control of volume, which is generally a real problem in a.c. tube receivers.

A short aerial was used with the set since there are 35 turns for coupling to the antenna in the first r.f. transformer. If a long antenna is used, a fixed condenser C_1 of .0001 or .00025 mfd. should be cut in series with the aerial.

In lining up such a receiver, all voltages should be checked with the volume control set up towards maximum value. There is nothing complicated about the procedure. The bias resistor R_1 can be adjustable for obtaining the best setting. R_1 and R_3 can be mounted either in the set or in the power pack.

No output transformer is shown since most good dynamic loudspeakers have an output transformer incorporated in them. Generally better results can be obtained by using a 30 to 50 henry choke to supply the plate voltage to the '71 tube and connecting the loudspeaker output transformer from filament to plate through a 4 mfd. condenser. This prevents the d.c. plate current from lowering the primary inductance of the output transformer and so dropping some of the lower notes.

The main idea of this article is to encourage the use of the screen-grid tube as a power detector in a.c. receivers. The choice of r.f. amplifiers is immaterial as long as sufficient gain is obtained. For local reception only, one stage of screen-grid r.f. amplification is generally sufficient with this system. The type of power tube is immaterial also, either a '71 or '50 being satisfactory. The '50 tube will, of course, give much greater volume without distortion, but costs more and requires a more expensive power pack.

Radio Picture Transmission and Reception

Photoelectric Equipment and Methods for Visual Communication

By JOHN P. ARNOLD, *Departmental Editor*

AMPLIFICATION OF PHOTO-ELECTRIC CURRENTS

IT is largely due to the introduction of vacuum tube amplifiers into telephony that the transmission of images has become something more than a rather uncertain and certainly an elementary form of communication. Without them, phototelegraphy and television would have very little significance in the modern sense, since the electrical output of the alkali metal photoelectric cell, which is the heart of practical systems, is extremely minute and, without amplification, such currents cannot be transmitted over ordinary communication channels.

The employment of the thermionic vacuum tube itself in amplifiers is quite familiar to those engaged in communication engineering. It is enough to say that the tube acts as a relay in that small voltages delivered to its grid produce large variations in the plate circuit cur-

rent. With a number of these tubes and their attendant apparatus connected in cascade, variations of potential impressed on the grid of the first tube are increased in magnitude through the amplifying system to the point where the power developed in the output of the last tube is sufficient to be transmitted to a distant receiving station.

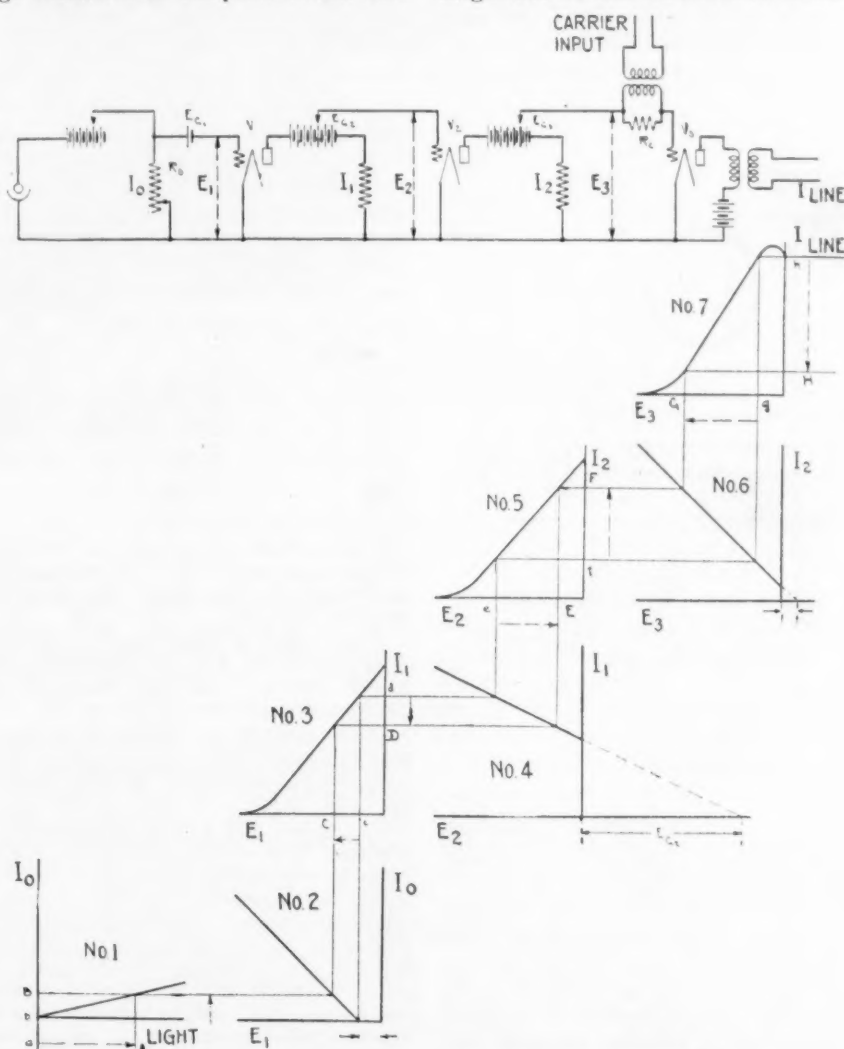
When a photoelectric cell of the alkali metal type is connected to the input of a vacuum tube amplifier, "the cell may be considered for our purposes as an impedance, the value of which is determined by the quantity of light reaching it. With no illumination this impedance is almost entirely a capacitance of the order of 10 m.m.f. When the cell is illuminated, this capacitance becomes effectively shunted by a very small conductance which is roughly proportional to the square of the voltage between the electrodes. For a fixed potential the magnitude of this conductance is nearly

a linear function of the illumination. With a suitable potential in series with the cell, then, there is obtained a current whose amplitude is proportional to the quantity of light reaching the cell." (See reference 1.)

For an example of how a cell is coupled to the input of a tube refer to Fig. 1, which shows a schematic diagram of a direct current amplifier used for the transmission of photographs over telephone lines. Below each individual circuit in this drawing are the characteristic curves of the tubes showing how the linear function between the illumination and the current is maintained during the progress of the signals through the amplifier. The current generated in the output of the photoelectric cell (at the extreme left) passes through the high resistance (I_0), and the potential tapped off this, of about 30 or 40 millivolts, is applied to the grid of the first vacuum tube. By appropriate grid biasing voltages (E_g), the range of current variations, is confined to the straight portion of the tube characteristic. The result of using the straight portions of the tube characteristic is that the received current, instead of varying between zero and a finite value, varies between two finite values. As will be shown elsewhere, this electrical bias can be matched by a mechanical bias of the light valve aperture at the receiving station.

When the same instruments can be used for both phototelegraphy and television, they are required to operate under much more rigorous conditions in the latter type of communication than in the former and are often taxed to the limit of their capacity. This is also true in the case of amplifiers. To transmit what we would consider a very crude series of images requires that not only a wide band of frequencies must be equally amplified over the entire range, but what is more difficult to encompass, the amplifier must operate from an input which is naturally small, since the light reaching the cell is that reflected from the subject to be transmitted.

In regard to the frequency range, to take a specific case, that of the Bell Laboratories system, it was found that a 50-line image transmitted at a rate of sixteen pictures per second required a frequency band from 10 to 20,000 cycles wide. By experiment, it was determined that the amplitude frequency must be constant to about plus or minus 2 T. U. and that the phase shift throughout the range must be "maintained so that the slope of its characteristic as a function



of frequency is constant to plus or minus 10 or 20 micro-seconds over all but the lowest part of the frequency range," where fifty times this limit was considered the maximum permissible. (See reference 2.)

The frequency band requirements as recounted in the foregoing paragraph were readily met, but the extreme weakness of the photoelectric currents demanded that especial precaution be taken to preserve the character of the signals. The usual sources of interference that arise in a case of this nature, where extreme amplification is necessary, are electrostatic and electromagnetic induction and mechanical and acoustic vibrations. The electrical disturbances can be reduced to a minimum by placing the amplifier close to the photoelectric cells, keeping the leads short and direct to avoid pick-up and closed loops of appreciable length. Vibrations, set up either directly through the supports of the apparatus or by sound waves acting upon the elements of the vacuum tubes, are avoided by lining the amplifier box with sound absorbing materials and by placing the tubes, wrapped in felt, in lead containers which are mounted on elastic supports. The so-called "peanut" tube has been found to be rather free from microphonic disturbances, and was used for the input stages of the amplifier to be described.

The amplifier circuit is shown in

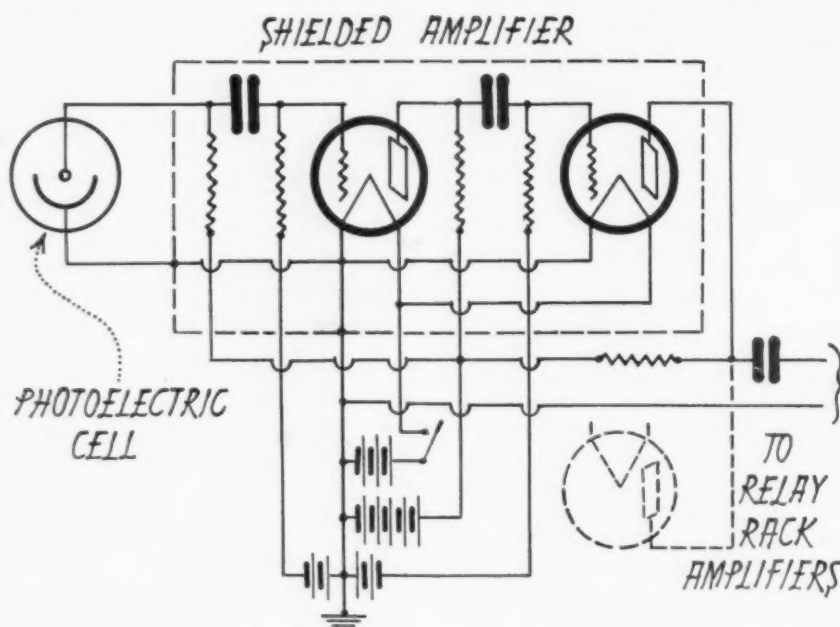


Fig. 2. Preliminary Stages of Television Amplifier Circuit

Fig. 2, or rather the preliminary stages of the amplifier, which is of the resistance-capacitance type and includes ten stages in all, the latter eight being similar to those shown in the diagram. The total amplification was about 130 T. U. It was from this known gain that the engineers were able to gauge the magnitude of the input to the amplifier from the photoelectric cell. This was found to be about 10^{-15} watts or, with a

100,000-ohm resistance in series with the cell, the potential available at the first tube was roughly 10 microvolts, which, needless to say, is very small indeed.

The utilization of the output of amplifiers at both the sending and receiving stations is most conveniently described, as it shall be, in another place; our purpose here has been merely to discuss the

tensity which are distributed by a scanning system (not shown) so as to reconstruct images of the original subject.

It is not necessary to describe the repeating or receiving amplifiers at length. Once the signals are raised to a manageable level they are not allowed to fall below the point where especially designed apparatus is again required. A

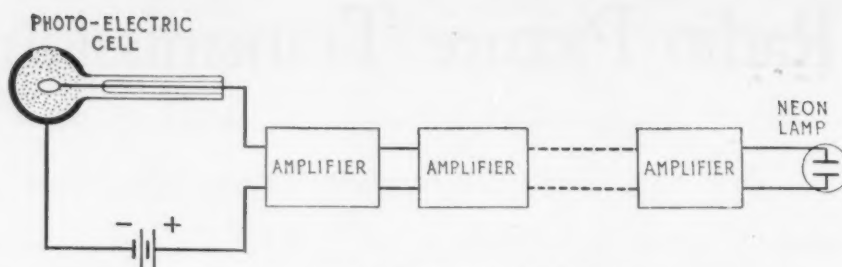


Fig. 3. Position of Amplifiers in Television Circuit

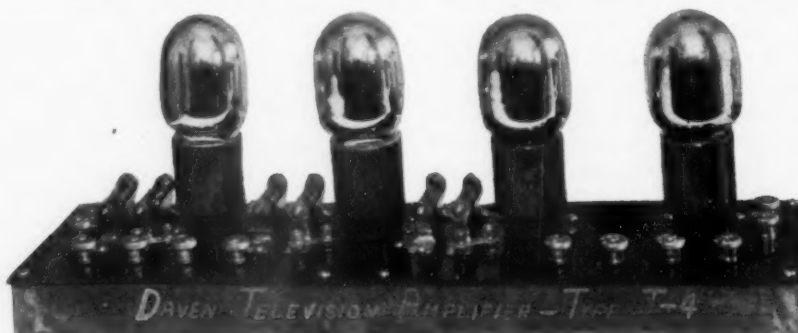


Fig. 4. Commercial Four-Stage Resistance-Coupled Amplifier of Received Television Signals

commercial type of amplifier, which may be used in conjunction with the ordinary radio receiver, is shown in Fig. 4. It is of the resistance-capacitance type. Particular care should be taken in designing this apparatus to avoid electrical and mechanical interference as in the case of the initial transmitting amplifier. Other disturbances than those already mentioned are usually traced to defective resistances used in the grid and plate circuits of the tubes. In other respects these amplifiers do not differ materially from those used in speech reception.

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ERRATUM NOTICE: In Fig. 2, page 22 of December, 1928, RADIO, the milliammeter connections are reversed. They should be changed so as to show a connection between the positive milliammeter terminal and the positive "milliamperes" input, and no connection between the positive milliammeter terminal and the negative milliammeter input.

AUTOMATIC SYNCHRONIZATION

The greatest drawback to present television technique is in the manual synchronization of the scanning disk of the receiver with that of the transmitter. The usual practice is to use a suitable variable resistance in series with the motor, together with a push-button that short-circuits the resistance for quick starting and for momentary acceleration of speed in getting the scanning disk in phase with the image. While such a method has served the purposes of experimental television, an automatic means of synchronization must be employed if television is to be placed in the hands of the layman.

Research has led to an automatic television unit, comprising a driving motor,

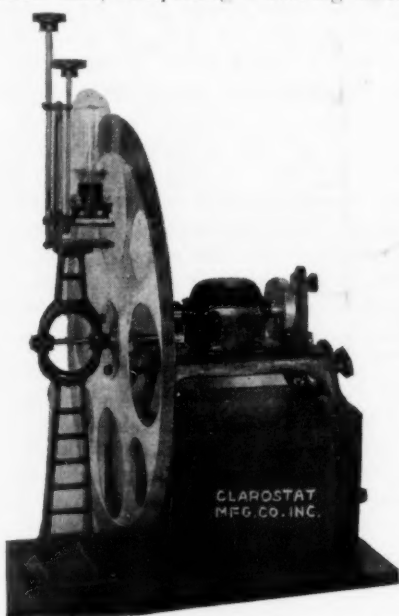


Fig. 1. Automatic Television Unit

automatic speed control, and an adjustable neon-lamp mounting, as shown in Fig. 1. The scanning disk is optional with the experimenter, since different television signals call for different scanning disks.

The motor is of 1/25th h.p., high-speed type, with suitable reduction gearing and flexible coupling for driving the scanning disk. This arrangement provides the greatest degree of correction, and eliminates troublesome alignment of drive shaft and scanning disk arbor. On the motor shaft is mounted a special centrifugal governor actuating a pair of contact points in series with the motor current supply. The contact points are shunted by an adjustable resistance. Thus when the speed of the motor rises above a given point, the centrifugal governor opens the contact points, thereby reducing the current to the extent of the resistance then thrown into the circuit. As the speed of the motor falls below a given point, the contact points close, short-circuiting the resistance and restoring full current to the motor.

Two power rheostats are used as the controls for obtaining any desired speed and also for trimming the sparking at the contact points by varying the degree of short-circuited resistance. The knobs of the rheostats are at the front end of the motor base. In addition, there is a small knob for adjusting the contact points, which serves as a vernier in obtaining precise speed even to one revolution.

Provision is made for mounting any scanning disk on the drive shaft. The neon lamp is mounted on an adjustable platform which, by means of rack and pinion movements, may be micrometrically raised or lowered as well as shifted from side to side. In this manner the light source can be readily adjusted for any variation in the neon lamp, as well as to align the image with the opening through which the image is visible.

In operation, the speed control knob is adjusted for the necessary speed, which may be anything from a few revolutions to many hundred. Vernier adjustments to within a single revolution or less, are made with the small knob that regulates the contact points. Finally, sparking at the contact points is reduced by the knob controlling the short-circuited resistance until it is almost imperceptible. A constant speed is now maintained with speed correction at every fraction of a revolution. The sparking at the contact points in no way interferes with the delicate receiver employed for television signal reception.

With this device, synchronization is simple and positive, and we might say that it is practical, at least within the limitations of our present technique. The television image is brought on the screen by varying the motor speed until the correct speed is obtained. This speed is then held just so long as the controls are left alone.

THE RAYFOTO SYSTEM OF STILL-PICTURE RECEPTION

(Continued from January issue)

THE mechanical printer unit consists of a metal cylinder around which is wrapped a sheet of photographic paper, a metal stylus or needle which rests lightly upon this paper, and a friction clutch whereby the rotation of the cylinder may be arrested while the driving machinery is still free to turn. This unit is mounted on a triangular base which can be placed over the turntable of any phonograph so as to drive the cylinder and cause the stylus to trace a spiral course over the photographic paper.

Three connections are necessary to unite the amplifier-oscillator circuit to the printer unit. Flexible leads from the binding posts *L* and *M* on the former are connected to the windings of a trip magnet located under the cylinder of

the printer. The only other connection is a wire from the high frequency end of the corona coil, *N*₁, to the stylus holder on the printer unit, Figs. 2 and 3.

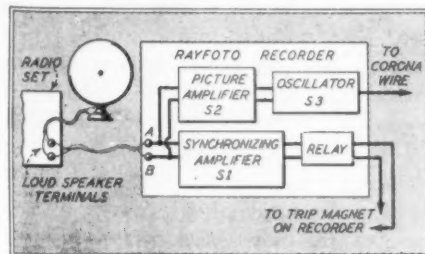


Fig. 2. Pictorial Diagram of Amplifier System

This lead should be of No. 22 or 24 wire and should not be longer than 3 ft., preferably about 1 ft. The wire should not be placed near any other conductor.

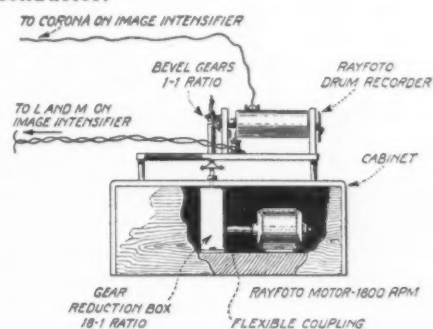


Fig. 3. Mechanical Printer Unit

Synchronism is effected by the "stop-start" system and is entirely controlled by the transmitting station. When the receiving cylinder makes one complete revolution, the check comes in contact with a lever on the trip magnet which is not released until a strong 1500-cycle signal from the transmitter, only passing through the synchronizing circuit due to its selective filter, actuates the relay and causes the trip magnet to release the cylinder. Thus synchronism

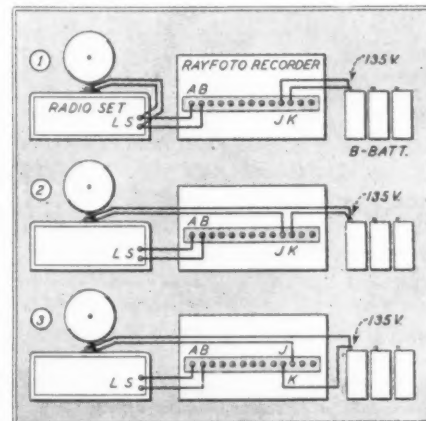


Fig. 4. Test Connections

is checked at every revolution or about 80 times for every linear inch of the picture.

After the various parts have been assembled and the batteries connected, the

constructor should first test the synchronizing circuit. To do this, the switch on the panel of the amplifier-oscillator unit is turned to the "On" position and the loudspeaker leads transferred to the points shown in diagram 2, Fig. 4. The receiver may be tuned to any station broadcasting speech or music. The input control R_1 is turned to the mid-position and the relay adjusted until it clicks whenever a loud signal of 1500 cycles is heard in the speaker. Once this adjustment of the relay is accurately determined, no further attention should be necessary. Any movement of R_1 to the right will serve to increase the strength of the signals. Such results indicate that the synchronizing circuit is in working order and will operate on actual picture broadcasts.

To test the picture amplifier, the speaker leads are connected as in diagram 3, Fig. 4. Musical notes, probably badly distorted as before, will also be heard in the speaker, but this indicates that the signals are being carried through to the oscillator circuit. When this circuit is functioning properly, a flashing light will appear in the indicator at the top of the corona coil. A small piece of white paper is now placed on the cylinder with the stylus resting lightly upon it and the control R_1 turned to the left so that the relay does not release the cylinder. The variable condenser VC_1 is adjusted until the stylus begins to discharge upon the paper. This spark will just be visible in the dark when R_1 is turned fully to the left. The intensity of the discharge varies at different settings of the condenser and the ideal adjustment is just below the point which gives the most intense spark. Once made this adjustment need not be changed. With these tests completed the receiver is ready for operation on picture signals.

From what has gone before the reader has probably rightly suspected that a dip into photographic processes will be required for the reception of pictures. It is for this reason: The light and shade of the transmitted photograph is converted to variations in intensity of an 800-cycle note which, after reception and amplification, is fed into the oscillator circuit where the high-frequency output of the corona coil causes the spark discharge from the point of the stylus. The ultra-violet light of this spark affects the photographically-sensitive paper wrapped around the drum in accordance with the tone values of the original picture. The reproduction is of course a latent image, (See "Radio" for Sept., pp. 25-26), that is, it must be developed and "fixed" in order that the picture may be seen. This is a very simple process and will give the experimenter no difficulty. The method will be described in the next article on the actual reception of pictures.

(To be Continued)

"TELEVISUALIZATION" AS IT WAS

The New York *Herald* of April 8, 1910, contained the following remarkable interview with William Vincent Pruscino, a young Italian, of Rochester, New York, who invented a gadget which he called a "televisualizer."

"The invention which I have made and the apparatus which I have succeeded in putting together is sufficient for me to make the assertion that I can visualize an object at a distance of thousands of miles. The object can be seen with its properties of color effects, dimensions and movements and will be equal to the original with the exception that the object itself cannot be felt with the fingers.

"The actual demonstrations made by me not many weeks ago have completely satisfied me that seeing at the distance of ten or twenty thousand miles is no more an impossibility; but on the contrary, is absolutely possible."

The young inventor feared that someone would steal his idea and destroyed the apparatus he had constructed. Finally, it was reported, he found financial support and was about to apply for a patent. Whether he ever did so, I am not certain, but I do not recall ever seeing any patent papers in that name.

In the New York *World* of December 12, 1909, appeared the following: "A press dispatch from Mexico City less than three months ago reported that Alberto Sanchez, an electrical engineer, had invented a 'teleradioticon' which would transmit vision as well as voice over an ordinary telephone wire. The apparatus resembled a short opera glass attached to a battery, and it was declared by persons who witnessed them that the tests were successful."

"Devices for 'seeing at a distance,'" remarked a disillusioned paragrapher in the New York *World* for June 10, 1909, "are now almost as numerous as catarrh cures and just about as effective."

But there you are and here is the moral. Twenty years ago they had just as grand ideas about television as we have today. Villon was pleased to inquire: "*Où sont les neiges d'antan?*" Twenty years hence, we may resume the plaintive strain—where are the snows of yesteryear?

RADIO PICTURE STATIONS

Station	Location	Wave Length
WMCA	New York City	526.0
WTMJ	Milwaukee, Wis.	526.0
KMOX	St. Louis, Mo.	275.1
WJR	Detroit, Mich.	399.8
WFBL	Syracuse, N. Y.	333.1
KFEL	Denver, Colo.	267.8
KXA	Seattle, Wash.	526.1
CKNC	Toronto, Canada	516.9
KWCR	Cedar Rapids, Ia.	229.0
KFPY	Spokane, Wash.	247.9
KSTP	St. Paul, Minn.	205.4

HOW TO MAKE 30 K.C. I.F. TRANSFORMERS

By R. WM. TANNER

SATISFACTORY transformers for use in the intermediate frequency stages of a superheterodyne can readily be made by winding the proper number of turns on a wooden dowel drilled with a center hole for an iron core. These transformers will peak at about 30 kilocycles and may be adjusted by changing the capacity of the condenser shunted across the primary of the transformer in the last stage.

The winding forms are wooden dowels 1 in. in diameter and $1\frac{3}{4}$ in. long and drilled with a $\frac{1}{2}$ -in. hole through the center.

The type of tube determines the ratio of the number of turns between the primary for secondary. For type '99 tubes this ratio should be 2.2 to 1; for the '01-A type 2.9 to 1; for the '12-A type 4 to 1; and for the '22 type 1 to 1. The following description applies to the '01-A type.

To reduce the capacity coupling between the primary and secondary, and thus increase the gain per stage, an electro-static shield is placed over the primary winding with the ends about $\frac{1}{4}$ in. apart. This shield is grounded to the outside copper shield, which is, in turn, grounded to the negative filament.

The primary consists of 690 turns (for '01-A tubes) of No. 36 enamel wire either scrambled or wound in layers, directly on the wooden form, leaving about $\frac{3}{16}$ in. at both ends clear. Two layers of heavy paper and the electro-static shield are placed over this. The shield should be slightly longer than the windings as described below. It might be well at this time to solder a short length of wire to the shield and then wrap on a few layers of paper. The secondary consists of 2000 turns of No. 36 enamel and may, like the primary, be wound either in layers or scrambled. After the windings are completed, "boil" each transformer in paraffine to exclude all moisture and to hold the wire in place.

Two disks of $\frac{1}{8}$ -in. cardboard or fiber, a little larger in diameter than the windings and having a hole in the center are tightly fitted on the ends of each wooden form. As much No. 36 soft iron wire as possible is then crowded in the holes drilled in the forms. It is a good plan to tie the wires together with fine silk thread to insure solidity of the core.

The output transformer has a primary of 375 turns of No. 36 enamel and secondary of 2100 turns. An electro-static shield is used in the same manner as in the other transformers. When connected in circuit, the primary is shunted by a .007 mfd. condenser.

A short length of insulated flexible wire should be soldered to the four leads

(Continued on Page 34)

The Screen-Grid R. F. Amplifier

By J. R. NELSON

Engineering Department, E. T. Cunningham, Inc.

ALTHOUGH the screen-grid tube is very valuable as a radio frequency amplifier, there are two widely-accepted misconceptions as to its performance. One is that the use of this tube results in a much less selective r.f. circuit than may be obtained with a three-element tube. The other is that it entirely eliminates the feed-back capacity between the tube elements, and consequently that shielding is not necessary to prevent oscillation. Let us consider these two misconceptions.

The selectivity S of a coil is defined by $S = \frac{\omega L_2}{R_2} = \frac{f_r}{f_1 - f_2}$ (1), where

L_2 is the inductance of the coil in henries, ω is 2π times the frequency, R_2 is resistance of the tuned circuit, f_2, f_1 are the frequencies at which the current has 0.707 of its value at resonance, and f_r is the resonant frequency. This equation defines the selectivity of the coil L_2 alone.

When the L_2 is used in an amplifier as shown in either Fig. 1 or Fig. 2, the

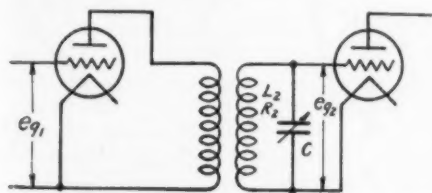


Fig. 1. Transformer-Coupled R.F. Amplifier Circuit

selectivity of the circuit is less than the selectivity of the coil alone. In Fig. 1 the plate resistance adds a resistance to the secondary resistance. The selectivity of the circuit is given by

$$S = \frac{\omega L}{R_2 + \omega^2 M^2} \quad \text{..... (2)}$$

for Fig. 1 and by

$$S = \frac{\omega L}{R_2 + \omega^2 L^2} \quad \text{..... (3)}$$

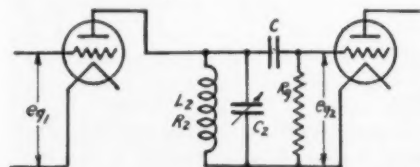


Fig. 2. Impedance-Coupled R.F. Amplifier Circuit

for Fig. 2, where M is the mutual inductance. In both circuits there is an additional loss caused by the presence of the condensers, shields, and any other metal that might be present. Equations 2

and 3 will be further discussed with the amplification equations.

In both Figs. 1 and 2 the voltage amplification, A_v , is defined as the ratio of e_{g2} to e_{g1} . The amplification is taken as the true repeater value, that is, the output does not affect the input so that regeneration is not considered.

The maximum possible amplification of a given tube and coil arranged in a circuit similar to Fig. 1 is given by

$$A_v = \frac{1}{2} \frac{\omega L_2}{\sqrt{R_2}} \frac{\mu}{\sqrt{r_p}} \quad \text{..... (4)}$$

where μ is the amplification factor of the tube and r_p is the internal impedance of the tube. The coupling required to give this value of amplifications is

$$\omega^2 M^2 = R_2 r_p \quad \text{..... (5)}$$

Under these conditions the selectivity of the coil and circuit is only one-half of the selectivity of the tube alone. It is possible to obtain greater than optimum coupling with three element tubes excepting possibly the high μ tubes such as the CX-340. It is, however, impossible to obtain optimum coupling with screen grid tubes such as the CX-322 with the conventional radio frequency type of tuned circuits. The selectivity of a circuit using a screen grid tube is always greater than one-half of the selectivity of the coil alone.

If the coupling does not have its optimum value the amplification is given by

$$A_v = \frac{\mu}{r_p} \frac{\omega M}{R_2 + \omega^2 M^2} \quad \text{..... (6)}$$

In the usual three-element tube circuit ωM is considerably smaller than the optimum value (Eq. 5). The selectivity for this case has some value between the selectivity of the coil alone and one-half of its value. Equation 6 also applies to the screen grid tube. For the same selectivity considerably more amplification may be realized with the CX-322 tube than with a three-element tube, provided the coil is correctly designed.

In order to make ωM in (Eq. 6) large, a large primary inductance closely coupled to the secondary inductance L_2 is required. If the primary has about the same inductance as the secondary and is wound with rather large wire, about No. 24, the dielectric and eddy current losses will be excessive. The selectivity will be poor and the set will tune broadly. The correct method of obtaining a high primary inductance is to wind it with No. 36 wire or smaller. This may be coupled rather closely to the secondary without adding much loss. If the secondary coil is good, the value of the selectivity will be high.

As it is impossible to make M large enough with the conventional tuned radio frequency circuit to satisfy Eq. 5, the circuit of Fig. 2 is sometimes used. Its impedance at resonance is considerably higher than that of a circuit such as Fig. 1. The value of R should not be too low. It adds loss to the secondary resistance and the value of this added loss is

$$R = \frac{\omega^2 L^2}{R_g} \quad \text{..... (7)}$$

The higher the value of R_g , the smaller is the added loss. The amplification is given by

$$A_v = \frac{\mu}{r_p} \frac{\omega L_2}{R_2 + \omega^2 L^2} \quad \text{..... (8)}$$

where $R_2' = R_2 + \frac{\omega^2 L^2}{R_g}$

This amplification is higher than may be obtained with the usual transformer-coupled circuit. The selectivity is poorer, however, with impedance coupled circuits. Almost as much amplification may be obtained with a transformer-coupled circuit, so that, in general, an efficient transformer-coupled circuit may be satisfactorily used with a screen grid tube.

In any circuit using a tube with a control grid and plate the input and output circuits are coupled together through the control grid-to-plate capacity. This mutual coupling is enough to make most circuits oscillate in broadcast frequency amplifiers. There have been several methods of stopping the oscillations. The first one used was that of adding loss to stabilize the amplifier. This has not been very satisfactory except for the one using a resistance of several hundred ohms in series with the grid. The best method before the screen grid tube was available was to add another condenser in a bridge network to neutralize the effect of the control grid-to-plate capacity.

The plate and the control grid in a screen grid tube have a grounded conductor, the screen grid, to shield them. This prevents any lines from one reaching the other. If the screening or shielding action were perfect, the control grid-to-plate capacity would be zero. The screening is very good, however, in a tube such as the CX-322. The control grid-to-plate capacity is only .025 mmfd., less than one-hundredth of that of an ordinary three-element tube. The capacity is not increased by socket connections as the control grid is brought out at the top of the tube and the plate at the bottom.

An Effective Antenna Coupling Unit

By THOS. A. MARSHALL*

The limit of stable amplification of a single stage amplifier using the CX-322 plotted against frequency is shown by Fig. 3A. The value of control grid-to-plate capacity used in calculating this curve was .025 mmfd. This is the maximum permissible value and most tubes have a capacity under .02 mmfd. This is for an extreme condition so that the permissible amplification in practice would be higher than shown by the curve.

Fig. 3B shows how the critical value of amplification would be reduced when the total control grid-to-plate capacity is doubled. This condition might be found when no shielding is used between input and output circuits. From the reduction of the critical value of amplification to cause oscillations to occur it is easily seen that care is required in shielding and arrangement of the parts if a stable amplifier is desired.

To show what has been accomplished in the screen grid tubes, Fig. 3C shows the limit of stable amplification using a CX-301A tube in a single stage tuned radio frequency amplifier. The limit would be raised if the grid plate capacity were neutralized, or if a resistance were inserted in series with the grid. Curves A and C, however, are for the same conditions and show how much the screening raises the limit of stable amplification, which is the factor of most practical interest in the operation of the screen grid tube.

Although the critical value of amplification or slightly greater amplification may be realized in practice, the circuit might or might not oscillate, depending

(Continued on Page 34)

THE antenna coupling unit described in this article was designed primarily to prevent re-radiation of the oscillating energy developed by the autodyne detector circuit and to aid in reducing interference to nearby receivers. After a suitable circuit was developed to bring about the required results it was found that a resultant increase in sensitivity and selectivity was made possible. The signal was increased about three

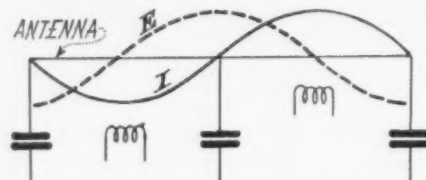


Fig. 1. Current and Voltage Distribution in Antenna System

times in the 18,000 k.c. band and was estimated to be as high as fifty times in the lower bands.

It is the writer's belief that most radio fans pay little attention to the problem of obtaining the most out of their receiving antenna since practically all receiving circuits are coupled to the antenna through a small condenser or by means of a few turns of inductance.

Fig. 1 shows the voltage and current distribution in an antenna system. For maximum signal strength the type of

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coupling along the antenna should be used as shown. Note that for greatest signal strength the antenna should be coupled capacitively at the voltage points and inductively at the current loop points. Thus, in an antenna operated at full wave or second harmonic, the antenna turns are at the current node point with maximum voltage in the coil system, resulting in the transfer of energy taking place capacitively rather than by inductive coupling. At other points the current and voltage distribution in the antenna system may be such that their respective fields neutralize one another, resulting in practically zero coupling.

Fig. 2 shows the distribution of voltage and current where an antenna is $\frac{3}{4}$ wavelength and adjusted for reception on the third harmonic. Fig. 3 shows the voltage and current distribution

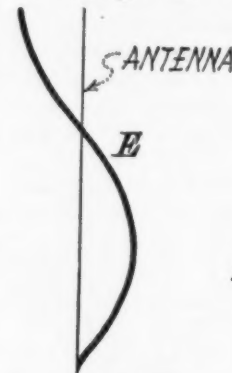


Fig. 2. Distribution in $\frac{3}{4}$ Wavelength Antenna



Fig. 3. Distribution in Full Wave Antenna

taking place in an antenna operated on full wave or second harmonic. Note that the antenna turns are at the current node point with maximum voltage in the coil system.

In all antenna systems, the current and voltage nodes and loops move along the antenna system as the frequency of the oscillating detector circuit is varied.

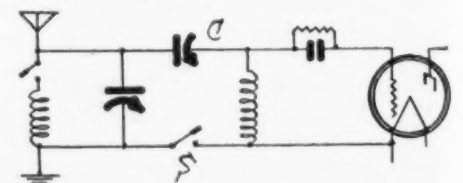


Fig. 4. Adjustable Antenna-Coupling Circuit

Thus, it is at once understood that a condenser-coil combination as illustrated in Fig. 4—will enable a variation of voltage and current distribution along the antenna system to be adjusted at will, resulting in a reduction in losses of signal strength taking place between current and voltage nodes as experienced in untuned antenna systems.

The greatest impedance of the con-

(Continued on Page 36)

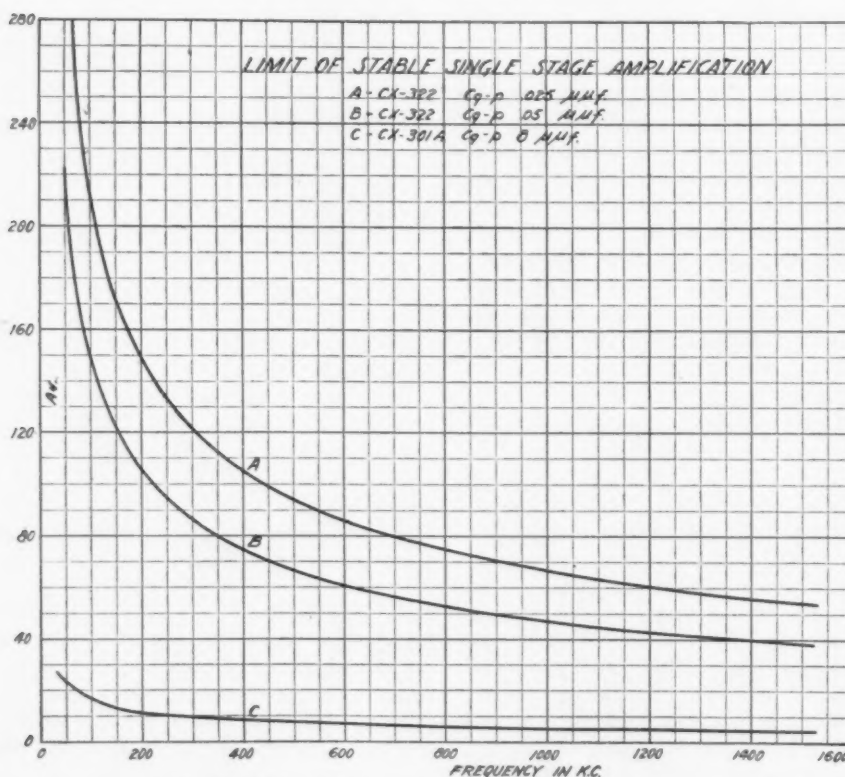


Fig. 3. Limits of Stable Amplification with Single Stage



TECHNICAL BRIEFS



OVERLOADING of a '71 power tube, even at low volume, is frequently due to too low an effective plate voltage and may be corrected by using a plate-supply unit, which will deliver 180 volts to the plate of the tube, 40 volts negative to the grid, and 15 volts, more or less for the voltage drop through the output choke or transformer. This requires an output of say 235 volts as compared with the 160 volts delivered by some power packs. If the desired voltage is not available, a dry-cell battery may be used to supply the grid voltage.

AUTOMATIC regulation of the 110-volt a.c. supply for a.c. receivers may be attained by several methods. One is to use a special resistance unit or lamp in series with the a.c. line plug and power transformer, the resistance of the unit or lamp increasing with the greater heating effect of a higher voltage, thus preventing any overvoltage. Another uses a similar unit and a power transformer designed to have a large magnetizing current which is increased with an increase in line voltage if operated near saturation; the series resistance thus becomes more effective since it need not vary so much with temperature.

A third method utilizes a complex connection of special transformers and condensers which allow large magnetizing currents and leakage reactance. This scheme gives quicker control since the voltage control is independent of the time necessary for a series resistance to heat or cool.

MANUFACTURERS advise that the type '80 rectifier tube may be used with 350 volts on the plate instead of 300 volts at the same maximum current drain of 125 milliamperes. This means that this full wave tube may be used to operate a type '50 power tube at about 250 to 300 volts. At this voltage the power pack design is fairly simple and economical and the power output available is from 2 to 3 watts. Increasing the a.c. voltage 50 volts per plate means approximately an increase of 50 volts at the output to the radio receiver. This should be of good use for a new tube which is a "cross" between a '71

and a '50 power tube. With 250 to 300 volts available, a push-pull '10 or '50-tube amplifier may be used to good advantage for receivers or phonograph amplifiers requiring large output. The cost of the power pack for the same factors of safety should be cut down from one-third to one-half.

HEATER type a.c. tubes are easily used in r.f., detector and a.f. circuits where grid bias is obtained by means of resistances in the $-B$ lead. Ordinarily this lead is grounded and all returns such as r.f. coils and audio transformer secondaries connected to this lead. This allows easy wiring in a set if a metal chassis is used. The cathodes of the tubes are then connected through suitable resistances to ground and the plate current provides the necessary grid bias.

This makes the cathodes positive with respect to the grids, which is correct. For example, a "plate voltage" of 140 on the r.f. tubes would usually mean about 131 volts effective plate-to-cathode and 9 volts grid-to-cathode. In this connection, separate resistances may be used in each cathode lead with proper bypass condensers, or several cathodes may be connected together and only one resistance and bypass condenser are necessary. Incidentally, a fairly large bypass condenser should generally be used in order that there will not be too high an impedance for a.c. ripple currents.

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volts for a '71 tube. This is because the connection to the transformer winding midtop lowers the bias by one-half the filament voltage.

WHEN a loudspeaker is placed out of doors the low tones are not so loud as when the speaker is within a room. Out of doors the listener hears mostly the direct radiation, while indoors he also hears the radiation reflected from the walls and objects in the room. The amount of direct radiation from the average speaker is considerably greater at high frequencies than at low. As the auditory response is dependent upon the amount of energy which impinges on the eardrum, the low-energy, low-radiated notes are not so loud as the higher-energy high notes. Within a room more of the high-note energy is ordinarily absorbed, so that proportionately more low-note energy is reflected and heard. The tone of a loudspeaker is largely dependent upon the sound-absorbing characteristics of the room wherein it is placed.

THE operating principle of the condenser type of loudspeaker may be illustrated by connecting the output terminals of a radio receiver to the terminals of an ordinary condenser. A condenser is constructed with metal plates separated by paper and air, which act as a dielectric. Alternate plates are connected to the same terminal. When the rapidly changing alternating voltage is applied between the plates there is a correspondingly rapid change in the electrostatic attraction between them. This tends to compress and expand the whole condenser, so that if the bottom plate is fixed the top plate moves up and down, or backwards and forwards. This movement sets the air in motion and thus causes a sound whose pitch corresponds to the audiofrequency of the amplifier output. Such a condenser to produce loud notes must be large in size compared to the condenser ordinarily used in radio. Recent experiments with this type of speaker have been so successful that it will probably be on the market at an early date.

The limit of stable amplification of a single stage amplifier using the CX-322 plotted against frequency is shown by Fig. 3A. The value of control grid-to-plate capacity used in calculating this curve was .025 mmfd. This is the maximum permissible value and most tubes have a capacity under .02 mmfd. This is for an extreme condition so that the permissible amplification in practice would be higher than shown by the curve.

Fig. 3B shows how the critical value of amplification would be reduced when the total control grid-to-plate capacity is doubled. This condition might be found when no shielding is used between input and output circuits. From the reduction of the critical value of amplification to cause oscillations to occur it is easily seen that care is required in shielding and arrangement of the parts if a stable amplifier is desired.

To show what has been accomplished in the screen grid tubes, Fig. 3C shows the limit of stable amplification using a CX-301A tube in a single stage tuned radio frequency amplifier. The limit would be raised if the grid plate capacity were neutralized, or if a resistance were inserted in series with the grid. Curves A and C, however, are for the same conditions and show how much the screening raises the limit of stable amplification, which is the factor of most practical interest in the operation of the screen grid tube.

Although the critical value of amplification or slightly greater amplification may be realized in practice, the circuit might or might not oscillate, depending

(Continued on Page 34)

An Effective Antenna Coupling Unit

By THOS. A. MARSHALL*

THE antenna coupling unit described in this article was designed primarily to prevent re-radiation of the oscillating energy developed by the autodyne detector circuit and to aid in reducing interference to nearby receivers. After a suitable circuit was developed to bring about the required results it was found that a resultant increase in sensitivity and selectivity was made possible. The signal was increased about three

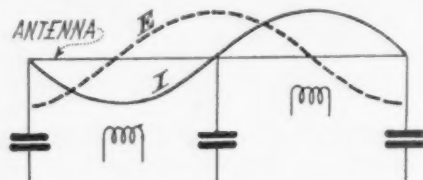


Fig. 1. Current and Voltage Distribution in Antenna System

times in the 18,000 k.c. band and was estimated to be as high as fifty times in the lower bands.

It is the writer's belief that most radio fans pay little attention to the problem of obtaining the most out of their receiving antenna since practically all receiving circuits are coupled to the antenna through a small condenser or by means of a few turns of inductance.

Fig. 1 shows the voltage and current distribution in an antenna system. For maximum signal strength the type of

* Formerly of the U. S. Naval Research Laboratory, Bellevue, D. C.

coupling along the antenna should be used as shown. Note that for greatest signal strength the antenna should be coupled capacitively at the voltage points and inductively at the current loop points. Thus, in an antenna operated at full wave or second harmonic, the antenna turns are at the current node point with maximum voltage in the coil system, resulting in the transfer of energy taking place capacitively rather than by inductive coupling. At other points the current and voltage distribution in the antenna system may be such that their respective fields neutralize one another, resulting in practically zero coupling.

Fig. 2 shows the distribution of voltage and current where an antenna is $\frac{3}{4}$ wavelength and adjusted for reception on the third harmonic. Fig. 3 shows the voltage and current distribution



Fig. 2. Distribution in $\frac{3}{4}$ Wavelength Antenna



Fig. 3. Distribution in Full Wave Antenna

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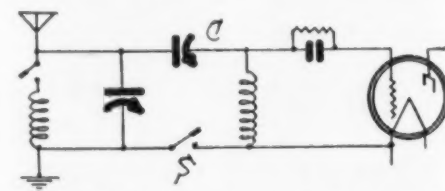


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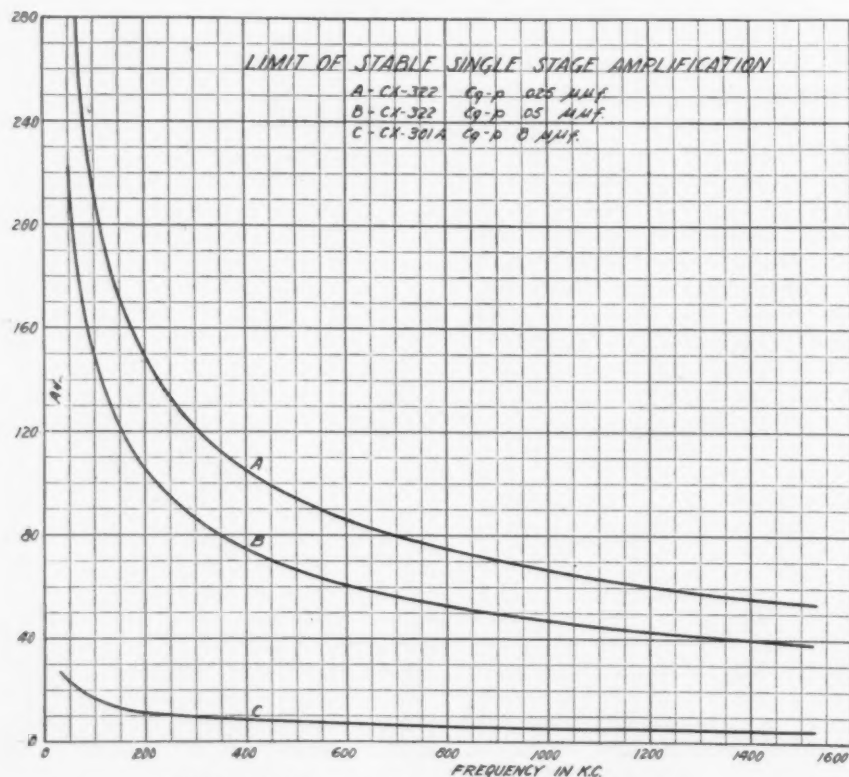


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A SIMPLE explanation of how grid detection is accomplished with a grid condenser S and a grid leak or high resistance R may be based upon the circuit diagram in Fig. 2. S readily passes an incoming small high-frequency a. c. voltage, which is blocked by R . Such a

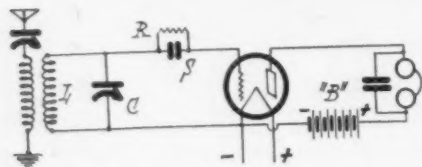


Fig. 2. Detection With Grid Leak and Condenser

voltage makes the grid alternately negative and positive.

When the grid is negative, its electrons are attracted in the external circuit through R to the relatively positive filament. This establishes a small grid current with a voltage drop across R , whose grid end is thus made more negative than its filament end.

When the grid is positive, no electrons are attracted through the external circuit to the filament and so there is no grid current. Current flows in the grid circuit only when negative half-cycles are impressed upon the grid by a carrier. Consequently, the carrier and its accompanying modulations are rectified or detected by the grid circuit. *Q.E.D.*

The rectified grid current consists of (1) the carrier high-frequency a.c. component, which is bypassed by S , (2) the modulating audio-frequency a.c. component which passes through R , and (3) a d.c. component which also passes through R . The resulting variations in grid potential control the internal flow of electrons from the filament to the positive plate, varying the intensity of the plate current, which thus reproduces and amplifies all the modulating variations in the rectified grid current.

PLATE detection in a vacuum tube circuit is accomplished with a C battery, which is used instead of a high-resistance R to keep the grid negative. If the grid be made sufficiently negative it will entirely cut off the plate current by preventing any electrons from reaching the plate. As the grid is made less and less negative, by decreasing the C voltage, more and more plate current flows.

Thus a series of measurements might show that .001 milliamperes flows at -5 volts C , .01 m.a. at -4 volts, 0.1 m.a. at -3 volts, 0.2 m.a. at $-2\frac{1}{2}$ volts, 0.3 m.a. at -2 volts, 0.5 m.a. at $-1\frac{1}{2}$ volts, 0.75 m.a. at -1 volt, 1.25 m.a. at $-\frac{1}{2}$ volt and 1.5 m.a. at zero volts. It will be noticed that between 0 and $-1\frac{1}{2}$ volts the current decreases from 1.5 to .5 m.a., a difference of 1 m.a., while between $-1\frac{1}{2}$ and -3 volts it decreases from 0.5 to 0.1 m.a., a differ-

ence of only .4 m.a. For the same difference in voltage less than one-half as much current flows when the voltage changes from $-1\frac{1}{2}$ to -3 volts as when it changes from 0 to $-1\frac{1}{2}$. This is due to the inherent characteristic of a detector tube.

When no signal is being received the grid bias is steady at say $-1\frac{1}{2}$ volts, likewise the plate current is steady at say 0.5 m.a. When the grid is being positively charged during a positive half cycle the plate current is very greatly increased. When the grid is being negatively charged during a negative half cycle the plate current is only slightly increased. For equal voltage changes on either side of the normal grid voltage, there are unequal changes in plate current, that for positive voltage changes being much greater than that for negative. Since the phones are actuated by the plate current, they easily respond to the large amount during a positive half cycle and do not respond to the small amount during a negative half cycle. Thus the plate current rectifies or detects the carrier and its accompanying modulating frequencies. The high frequency carrier oscillations cannot pass through the high inductance of the head phones but are bypassed by the condenser shunted across them.

AN ELECTROLYTIC rectifier consists of two dissimilar metallic electrodes suspended in a suitable electrolyte or solution. Its action depends upon the formation of a film on one electrode called the anode. This film conducts current much more freely in one direction than in the other and so rectifies an alternating current which is impressed across the two electrodes. When an a.c. voltage is applied, the free electrons readily pass from the anode through the film to the cathode, whereas the heavier positive ions with their greater mass cannot penetrate the film and consequently remain in the cathode.

The anode metals in ordinary use are aluminum, tantalum, or duralumin, the latter being an alloy consisting of 94.66 per cent aluminum, 3.93 per cent copper, 0.56 per cent manganese, 0.50 per cent magnesium, 0.33 per cent silicon, and 0.02 per cent carbon. Tungsten, bismuth and magnesium are among the other metals which may be used as anodes.

The cathode requires an inert metal which is not affected by the electrolyte. Lead, carbon and iron are frequently used.

The electrolyte depends upon the kind of electrodes. For aluminum and lead, either ammonium borate or ammonium phosphate are the best, the former handling heavier charging voltages but giving a sediment which does not occur with the latter. For the tantalum and lead, or lead peroxide electrodes, the electro-

lyte is dilute sulphuric acid. For duralumin and durion electrodes, the electrolyte is diammonium hydrogen phosphate, potassium dichromate and oxalic acid.

BETTER selectivity is secured by placing several tuned circuits ahead of the tubes in an r.f. amplifier than by using the same number of tuned circuits with one between each tube, provided there is no regeneration in either case. In the former case only the sharply tuned final wave form is amplified. In the latter case there is considerable amplification of the broader wave form passed by each successive circuit. Such a pre-selector is especially suitable for use with screen-grid tubes since they have little or no regeneration.

Difficulties in its use include that of obtaining an efficient untuned r.f. amplifier, its amplification of radio disturbances unless well isolated, and the difficulty of obtaining an efficient coupler. When regeneration is present, as with most three-electrode tube circuits, it not only increases the amplification, but also improves the selectivity. This regeneration usually more than offsets the damping due to the reflection of the tube plate impedance from the primary of the r.f. transformer into the tuned circuit.

MOST dry plate or so-called contact rectifiers use either copper sulphide or copper oxide disc or coating as a unit for unilateral conductivity. One of these discs offers a low resistance to the flow of current in one direction and a high resistance to current flow in the other direction. Each unit is capable of withstanding about 4 volts per element without breaking down, so that when higher voltages of alternating current are to be rectified or converted into pulsating direct current it is necessary to connect a number of units in series. The maximum safe current density per square inch of plate area is about .2 amperes. The copper oxide unit is claimed to have a rectifying ratio of 10,000 in one direction to one in the other, as compared to a 75 to one ratio which is ascribed to a copper sulphide unit. Rectifying efficiencies vary from 35 to 57 per cent and decrease with overheating or length of use, thus having to be eventually replaced to avoid hum. It is claimed that the copper sulphide elements deteriorate more rapidly than the copper oxide elements. For low voltage work these rectifiers are made in capacities up to 200 amperes.

A dynamic speaker, according to the R. M. A., "is one in which a portion of the conductor carrying the alternating signal current is a part of the moving system, the force producing the motion being due to the location of this conductor in a magnetic field."



The COMMERCIAL BRASSPOUNDER

A Department
for the Operator
at Sea and Ashore



Edited by P. S. LUCAS
R. O. COOK, Assistant



SOME time ago we discussed the comparative ability of the average foreign and American operators. A letter from Francis Bayley, brasspounder on the British ship *Adrastus*, gives an interesting slant on the radio situation in England. "The British schools are pretty bad," he says, "and the men can be divided into two classes; those who are keen on radio, and those who have been tried and found wanting in everything else." Well, there's a parallel case.

Mr. Bayley reminds us that it is more difficult to send fast on a "pump handle" than on a bug, which is very true. "Our instructions require full use of our call sign," Mr. Bayley goes on to say, "and there is often somebody ready to jump on us if we don't obey. In one company, at least, ops have to enter in the log the number of times the call sign is repeated each call."

To the American mind this super-adherence to regulations seems cumbersome and hidebound. Yet, where the Briton is perhaps over-cautious, the American may be too hasty. A happy medium might result in increased efficiency on all sides, but try and get it. Let's remember, at any rate, that the foreign operator may have to put up with all sorts of antique regulations which do not allow him to handle traffic in the U. S. A. style; and let's give him the benefit of the doubt before "putting him in his place."

There has been quite a rumpus between 2UO and the A. R. R. L. headquarters for a long time, and the consensus of opinion seems to have been that 2UO was in the right church but the wrong pew. Now, that a commercial license has been assigned this station everybody is happy, and the rare, snappy press that used to be sent on 40 meters is now to be found on 32. WHD is the new call.

SQUEEZING A RELAY OUT OF THE FLIVVER

By R. MADDEN, KDIY

Another of the many pieces of a Ford motor car that has found its way into the radio shack is the magnet coil of the horn. By mounting this on a board with an ordinary key as shown, it makes a fine keying relay. A piece of flat iron is mounted under

the knob of the key and 6 volts connected to the coil through the bug. The original will handle the bug at full speed with the weights removed.

THE TITANIC DISASTER

As Dug Out of the Old Congressional Records
By WM. A. BRENNIMAN, KOZC

The world has recently been stirred to the depth by a maritime disaster which almost parallels that related below. It is possible that the Vestris tragedy will result in a more liberal use of radio in time of peril. Perhaps the officers may be held accountable to the government at such a time rather than to a nickel-worshipping steamship company.

CQD CQD SOS SOS DE MGY TITANIC SINKING, PLEASE RUSH ALL POSSIBLE ASSISTANCE, RUSH, RUSH" were the few terse, electrifying words clicked out by the wireless operators of the gigantic White Star liner *Titanic* on the night of April 14, 1912. The startled world was thus given its first inkling of the terrible disaster that befell the doomed liner, and wireless, or radio, was skyrocketed to the attention of the world as a utility and safeguard of the utmost importance.

The *Titanic* was equipped with a 5-k.w. disk discharger, magnetic detector, valve receiver and emergency gear. It was the only vessel afloat that had one of the new disk discharger installations and it boasted a range of about 500 miles at daylight. At the key of the splendid liner were Jack Phillips, chief operator, and Harold S. Bride, 22, second operator. Both were in the employ of the British Marconi Company and were being paid a monthly salary of six and four pounds Sterling, respectively, (about \$28.00 and \$20.00 in American money), all for being entrusted with the safety of nearly 2500 souls in case of emergency.

On the morning of that fateful Sunday, trouble had been experienced with the insulation on the main panel of the transmitter and Mr. Phillips has arisen considerably earlier than usual to help repair the trouble. This finished, he started his 4 p. m. to 2 a. m. watch, with Mr. Bride's promise to relieve him at midnight, as the former was not feeling very well. At 5 p. m. the SS. *Californian* called the *Titanic*, MGY, and the *Baltic*,

MBC, with information regarding ice, stating that that ship had just passed three large bergs and a large number of growlers or smaller bergs. The *Baltic* acknowledged the call with the signal "RD," used at that time to QSL a message. The *Titanic* heard the report and copied it, but did not acknowledge immediately, as Mr. Phillips was working on his abstracts and had them all about him on his desk. After about twenty minutes, however, he gave the *Californian* the "RD" signal, whereupon the latter sent "TIS," or the finishing signal used at that time. Mr. Phillips took the information regarding the position of the icebergs to the bridge at about 5:30 p. m., ship's time, and the officer on watch figured they would be in the vicinity of the bergs about 11 p. m. that evening.

From 6 p. m. to 10 p. m. Phillips exchanged several regular messages with nearby ships and at 10 p. m. he listened in to press reports from the Cape Cod station, which called the *Titanic* at 11 p. m. with a large number of messages, which kept the two stations in communication up to the time of the collision. As closely as could be ascertained, this happened at 11:50 p. m., Sunday, New York time. There was just a slight grating and a little lurch of the vessel to port. The blow was so slight that it did not even serve to awaken passengers who had retired, although most of the ship's crew who were not on watch were awakened by the unusual lurch of the vessel and came on deck to investigate. No one thought, however, that anything serious had occurred.

Mr. Bride had planned to go on watch at midnight, so arose about 20 minutes before. The operators' sleeping quarters adjoined the operating room and before dressing, Bride stepped into this room, asked Phillips how he was making out and was told that there were still a number of messages for Cape Cod. Mr. Bride then dressed and put on the phones.

It was at this instant that the boat struck, but with such a slight shock that Phillips continued his preparations for retiring. Not long after that the captain came into the operating room and told Bride that he had better get assistance. Phillips, hearing him, came out into the operating room and asked the captain if he wanted him to use the distress call. The captain said he did, so Phillips took over the key and sent the "CQD" signal about half a dozen times, signing "MGY," the call letters of the *Titanic*.

The SS. *Frankfurt*, DFT, was the first vessel to answer the call. Phillips advised Bride that the *Frankfurt* had answered and asked him to take the information to the bridge. Captain Smith then asked for the *Frankfurt's* position. The *Carpathia* answered the call, giving her position, and said she was already coming to their assistance. Phillips then raised the *Olympic*, but while he was working him the captain came in and

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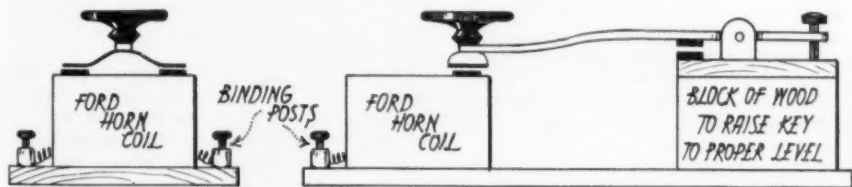


Fig. 1. Constructional Details of the Ford Horn Relay

With the Amateur Operators

PARABOLIC REFLECTORS

By A. BINNEWIG, JR.

THE gain in power radiated in a given direction with a simple parabolic reflector is of the order of 4 times that obtained with a single vertical wire favoring no particular direction. And because the expense entailed by the addition of the few wires and insulators necessary to this directional antenna and the installation of more powerful equipment to secure the same four-fold increase in result, the parabolic reflector is of great value to the transmitting amateur who must of necessity use low power. Parabolic reflectors, at the shorter wavelengths, are small enough to be conveniently used in most amateur stations. They are simple in construction and more effective than many of the more elaborate systems.

All the wires used in the reflector are approximately the same length as the main antenna and all are resonant at the frequency used. Untuned wires will not operate properly. For practical amateur use 3 or 5 wires arranged vertically along a horizontal parabola should be used.

The actual transmitting or receiving antenna is placed at the focus. The larger the aperture of the reflector around it; that is, the further its sides are extended, the sharper is the resulting beam. If it is possible to construct the reflector on the roof of the shack a vertical zeppelin feed may be used to the aerial at the focus. The reflector may be fixed for transmission in any particular direction, or some method of pivoting may be devised; the main aerial remaining fixed as the axis, with the wires in the reflector maintaining their relative positions.

With the design curves for the parabolas (see Fig. 1) no calculations are necessary.

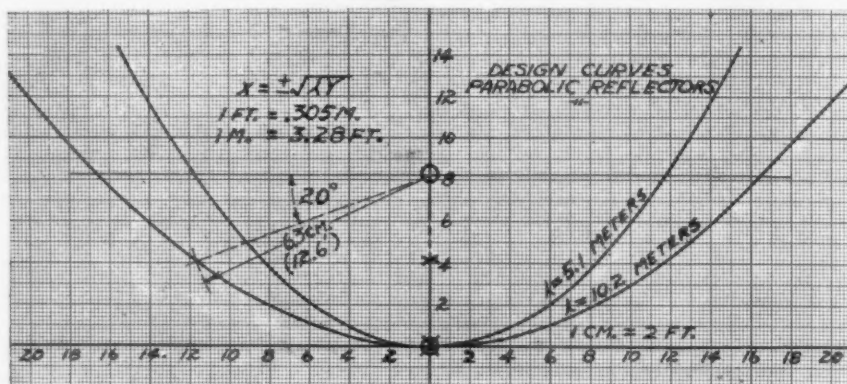


Fig. 1. Design Curves for Parabolic Reflectors

At the scale of 1 unit of the scale=2 ft. the parabolic curve may be plotted on the ground and the frame built to match. The spacing of the reflector wires should be one-half the length of one wave, and the radiating wire should be placed at the focus, which is one quarter of a wavelength from the vertex reflecting wire. In Fig. 1 the focus for the 10.2 meter parabola is shown by the circle 16 units or 8 ft. from the vertex, and the focus for the 5.1 meter wavelength is shown by the cross 4 ft. from the vertex. Best transmission will be in the direction along the axis from vertex to focus. This direction should be fixed by means of a compass.

A 3-wire 10-meter reflector would consist of 4 wires in all, each a half wavelength, or about 16 ft. long, the one at the focus

being coupled to the transmitter or receiver and the other three spaced along the parabola half a wavelength apart, with the center one at the vertex. The correct locations for other wires can be determined from the curves if it is desired to extend the ends of the parabola.

A 5-wire 5-meter reflector would consist of 5 wires spaced around the aerial according to the 5-meter curve. The shorter the wavelength used, the farther the sides of the parabola may be extended for a given amount of reflector space, and the sharper the beam. At $\frac{3}{4}$ meters a very sharp beam may be obtained and the whole system is so small as to be practically portable.

At $\frac{3}{4}$ and 5 meters, the wires used in the reflector should not be passed through the insulators and bent around as is cus-

tomary. It is better to bend back a 1-in. length at the end of each wire so as to leave a small "eye" which is strengthened with solder. This gives a definite length of wire for measurement. The wires are secured to the insulators with waxed cord.

Parabolas for any desired wave may be constructed by using the formula given in Fig. 1. X and Y are coordinates of any point on the curve and λ is the wavelength to be used. The three factors must all be in the same units; if λ is in meters, X and Y must also be in meters. By using the conversion factors shown, it is simple to change from one to the other.

The parabolic frame for the $\frac{3}{4}$ -meter reflector at 6BX, shown in the illustration, was constructed by bending two $\frac{3}{8}$ in. pieces of wood around finishing nails driven



$\frac{3}{4}$ Meter Parabolic Reflector at 6BX

into a parabola marked on the floor. These pieces were treated with hot water, bolted together and allowed to become dry before removal. It is only necessary to construct one frame for experimental purposes, for the wires can be accurately adjusted with a level or a plumb-bob anywhere around the frame. The bracing may be seen in the illustration. All parts are securely bolted together and the frame is quite sturdy. Two lengths of heavy cord should be tied diagonally across the front to securely brace the opening which is otherwise unbraced. Wire will not do.

Interference is reduced with directional transmission, and good efficiency for a given power is obtained. By the use of a reflector at both the transmitting and receiving stations, highly efficient and reliable communication is obtained.

FREQUENCY CONTROL PRECAUTIONS WITH A QUARTZ CRYSTAL

By H. A. CLARK, JR.

THE fallacy that a quartz crystal oscillator will operate only at the natural frequency of the crystal is apparently responsible for the fact that many crystal-controlled amateur stations are heard on other than the amateur bands. The chief factors, aside from the crystal's natural frequency, which affect the output frequency of a piezo-oscillator, are the working temperature, the load upon the unit, the plate potential of the oscillator tube, and the type of holder for the crystal.

The temperature at which the crystal oscillator will work may be controlled by suitable thermostatic units. But as the average amateur is not in a position to maintain units of this type, the next best bet is to calibrate the oscillator while working at normal room temperature.

The load imposed upon the oscillator as well as the potential upon the plate of the oscillator tube cannot be anticipated by the manufacturer of the crystal, so we can not expect his calibration of the crystal to fit our conditions.

The holder affects the frequency in three ways: First the supersonic waves imposed between the reflecting surfaces of the crystal and movable plate of the crystal holder; second, the capacity of the holder, and third, the mechanical load imposed upon the crystal by the movable plate of the holder.

(Continued on Page 42)

Inside Stories of Factory Built Receivers

RADIOLAS 60, 62 AND 64

THE R. C. A. Radiola 60 is a socket powered receiver employing an 8 tube superheterodyne circuit and a full wave rectifier. With the exception of the UX-280 rectifier tube and the UX-171A in the last audio stage, UY-227s are used throughout, and all plate, grid and filament voltages are supplied by the socket power unit which is a part of the set, built on a separate chassis but mounted in the cabinet at the right of the receiver assembly.

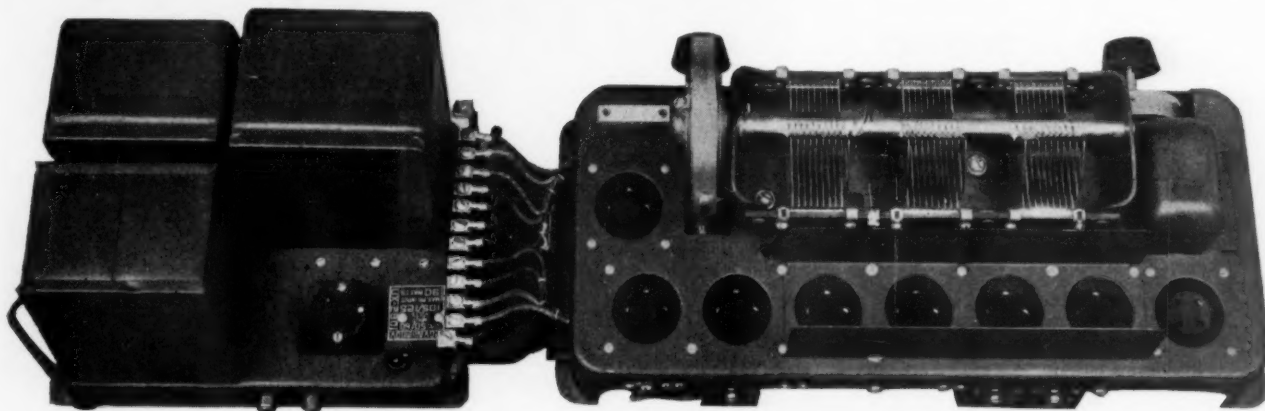
The first tube on the right of the chassis is an untuned stage of r.f. amplification which is coupled directly to the antenna and

grid to ground, across the *C* bias resistance as well as the secondary. The cathode is connected to the pickup coil of the oscillator, thence, with the oscillator cathode and that of the second detector, to terminal 5 and the center tap of the two filament transformers.

The two stages of i.f. amplification that follow the first detector are peaked at 180 k.c., and are neutralized by semi-variable condensers which are connected between their respective plates and the ends of the transformer secondaries. Trimmer condensers are also supplied for tuning or peaking the transformers at the correct intermediate frequency. The grid returns are taken out

on the left. The series and shunt trimming condensers are mounted on the rear of the chassis directly in back of the oscillator and second detector tubes. A 3000 ohm resistor and a fixed condenser are in series with the grid which is tapped into the coil at the center. 40,000 ohms are used as a leak resistance. Tube number 7 operates as a power detector, being supplied with a plate voltage of 160, and has an output sufficient to drive the power amplifier. A UX-171A tube is used in the audio amplifier stage.

Voltages for all circuits are obtained from a single power transformer with four secondaries. One secondary lights the filament of the UX-280 rectifying tube. An-



Chassis of Radiola 60

ground across a resistance of 2000 ohms. It is followed by a tuned r.f. stage which is neutralized by the small compensating condenser on the chassis at the rear of the first detector tube. The proper setting for this semi-variable condenser is just below the oscillation point and if it gets out of adjustment the sensitivity of the set will be lowered and oscillation may result.

The third tube is the mixer, or first detector. It is tuned by the middle condenser in the gang of three and is connected from

through the center taps of the secondaries. All i.f. trimmers and neutralizing condensers are mounted just underneath the tuning condenser assembly; the first i.f. transformer trimmer being on the right with the neutralizing condenser for the same stage just to the rear of it, the second i.f. trimmer and neutralizing condenser in the center and the two trimmers for the third i.f. stage on the left.

The oscillator is the next tube in the sequence and is tuned by the gang condenser

other, connected to 8 and 9 on the terminal strip, supplies the filament of the UX-171A power tube, and the third low voltage secondary provides voltage to all other sockets. This is connected to terminals 10 and 11.

The high voltage output is rectified and filtered and is divided between terminals 1, 2, 3, and 4, with the negative lead going to 7. This negative lead is grounded through a portion of the 450 ohm volume control potentiometer which is located at

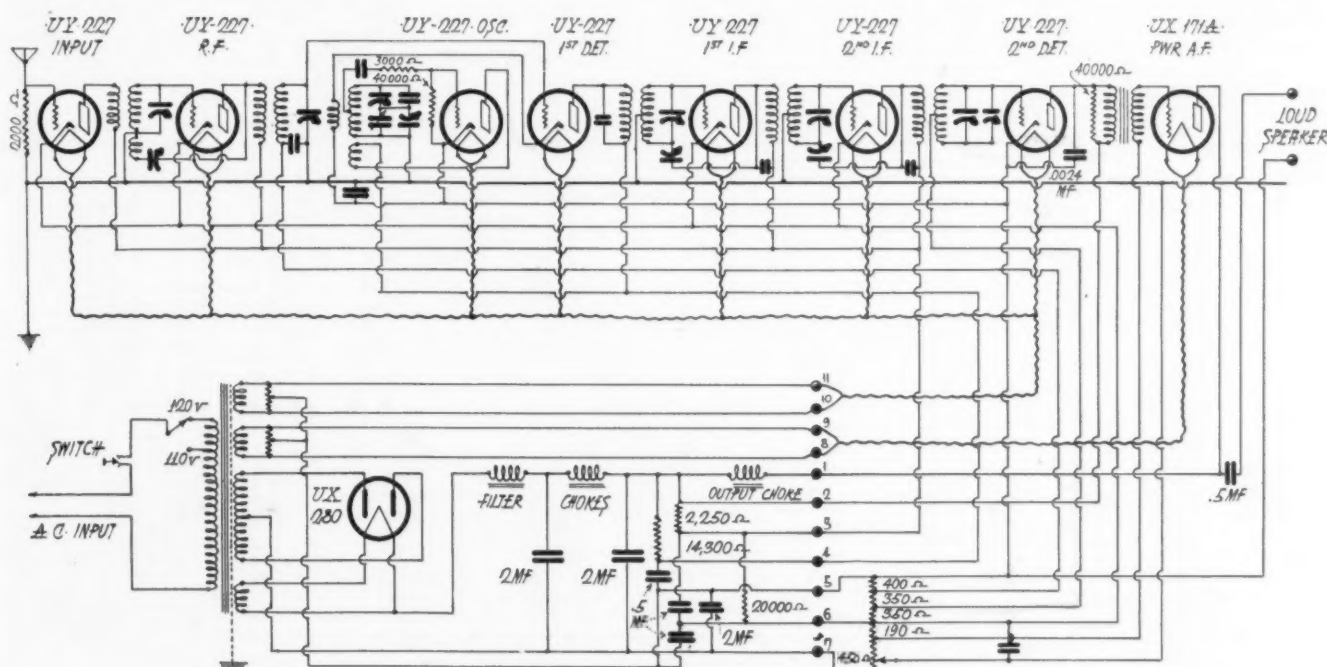


Fig. 1. Circuit Diagram of Radiola 60

the left end of the *C* voltage divider, just in front and below the level of the tuning condenser assembly. This potentiometer is connected in series with the *C* bias resistor and varies the grid bias on the two r.f. and two i.f. tubes. With the receiver in operation the voltage across terminals 1 and 7 should read 200. This includes the *C* bias voltage which is dropped through the 1290 ohms of resistance in the *C* voltage divider.

The voltage on the second detector terminals, 2 and 7 is 210, due to the fact that the output choke is not in the latter line, and the *C* bias is taken from this through 750 ohms of the divider. Terminal 3 is the positive lead for the r.f. and i.f. tubes, and should show 160 volts with respect to terminal 7, having been reduced to that figure by a 2250 ohm resistor in the power unit. The *C* bias in this case is obtained through the 190 ohm unit in the voltage divider plus the portion of the volume control between the cathodes and the ground. The plates of the oscillator and first detector tubes are connected to terminal 4, the voltage at this point having been reduced to 110 through the 14,300 ohm resistor in the power unit. The grid bias for the first detector is obtained from the 400 ohm tap and the oscillator operates with its grid at the same potential as the cathode.

Radiola 62 consists of the same receiving set as the 60, except that it is mounted on the upper shelf of a console cabinet with the power unit below. This set also includes a dynamic speaker mounted on a wide baffle-board, and incorporates a separate power unit for the speaker field.

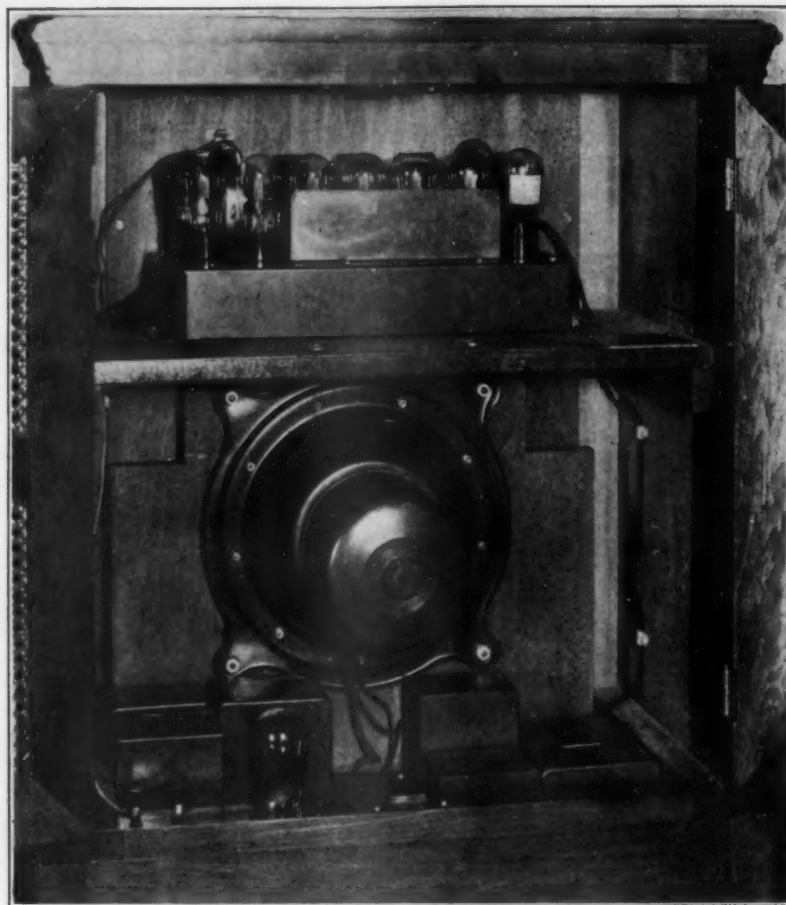
The Radiola 64 has several added features, chief of which is the automatic volume control. An extra UY-227 tube is employed for this purpose and is located between the second i.f. tube and the second detector. With maximum sensitivity, the volume knob is turned back until local stations come in with the desired strength only. Distant stations may then be tuned in with all their volume up to the same point without danger of a local overloading the tubes or the ears when passed. The meter on the panel indicates resonance to the frequency of an in-

coming signal, making the final resonance adjustment visual rather than audible.

The frame is not grounded but is connected to the cathodes of the r.f. and i.f. tubes. The ground goes to the grid of these tubes, being separated from the frame by the grid bias resistance.

A UX-250 power tube is used in the one and only audio stage, and the power supply must therefore be rectified by two UX-281s instead of the UX-280 full wave tube. The voltage divider for all circuits is directly across the rectified high voltage output, and

(Continued on Page 43)



Rear View of Radiola 62

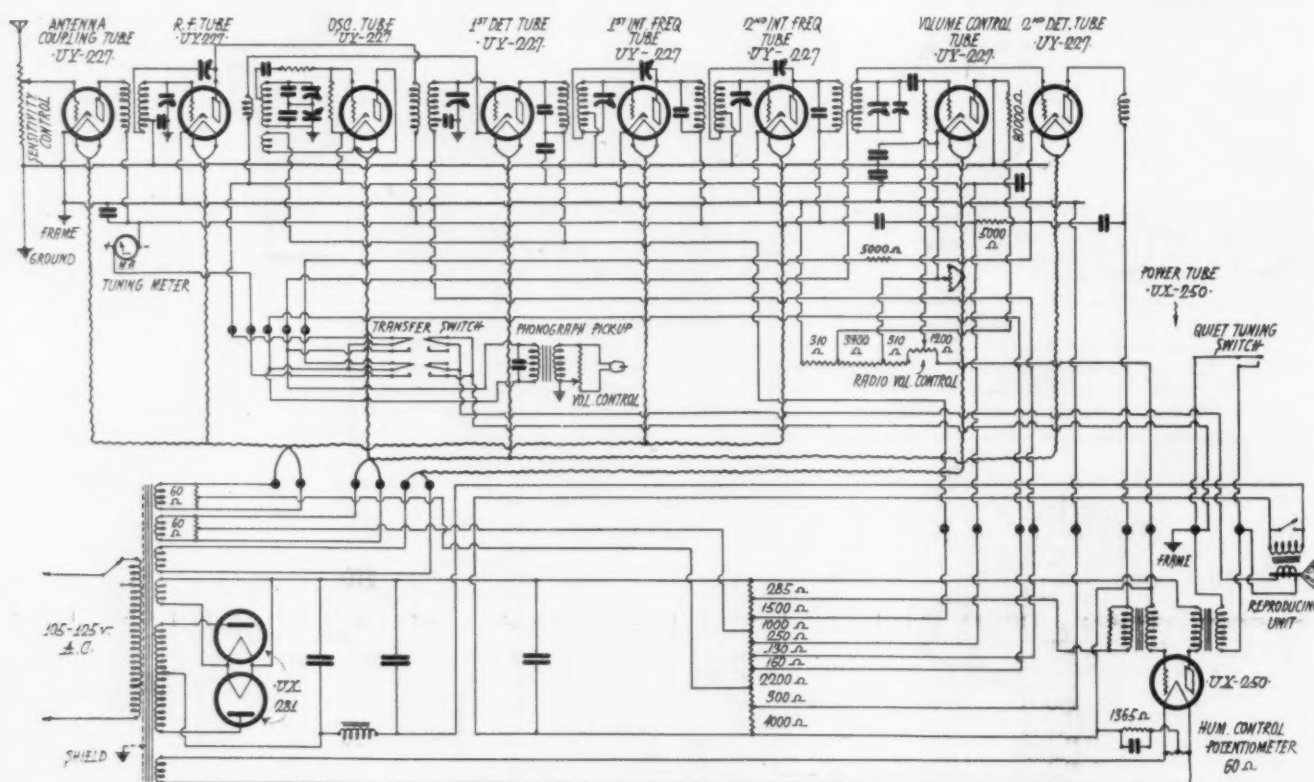


Fig. 2. Circuit Diagram of Radiola 64

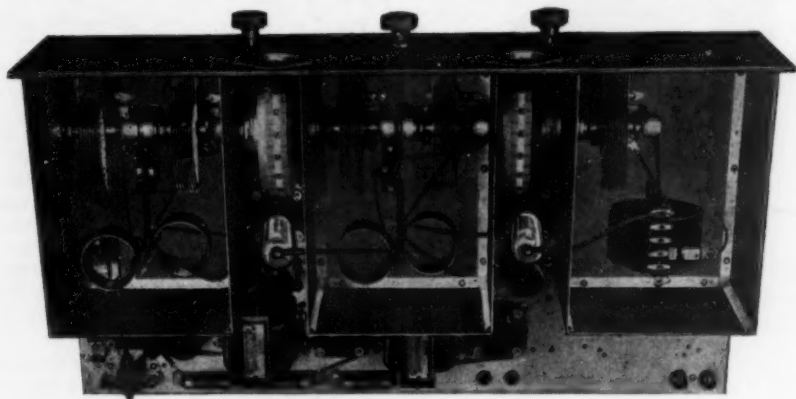
Radio Kit Reviews

HAMMARLUND-ROBERTS MASTER "HI-Q 29"

THIS kit is designed to facilitate the assembly of a high-grade receiver which has maximum sensitivity, selectivity and fidelity of tone reproduction with five d.c. tubes. Extreme sensitivity is secured from the great amplification provided by two screen-grid tubes in the r.f. amplifier. Great selectivity without cutting of side-bands is

ers is shown in the picture. The antenna coupling transformer, with its tapped primary, and its tuning condenser is in the left shielded compartment, the first r.f. transformer and its two tuning condensers in the center, and the second r.f. tuner in the right compartment, the two tubes and drum-dials appearing between the shield boxes.

These tuning circuits have been found by test to reduce an interfering signal which is 20 k.c. from resonance to 1/125 its original



Top View of Hammarlund-Roberts "Hi-Q 29"

attained by special tuned-grid and tuned-plate r.f. circuits which afford a band-pass filter. Great volume without distortion results from the use of high-grade audio transformers and a power tube.

The necessary high degree of selectivity without corresponding reduction in the high amplification of screen-grid tubes results from the five tuned circuits which are necessary to tune both the grids and the plates. This double tuning can be accomplished without oscillation from these tubes whereas ordinary tubes would thus be thrown into violent oscillation.

In each interstage r.f. transformer, the primary coil is connected to the plate circuit of the preceding tube and the secondary coil is connected to the grid of the following tube, the two coils being exactly similar and placed so as to have a coupling co-efficient of about 1 per cent. Each coil is tuned to resonance by means of a .00035 mfd. variable condenser, all of which are gang controlled from a single dial. The other dial controls the antenna coupler and its associated transformer.

The general arrangement of these r.f. transformers and their associated condens-

intensity while the top of the response curve remains nearly flat with steep sides. This form of response curve gives a minimum of side-band cutting.

The volume control consists of a 100,000 ohm potentiometer connected across the 45-volt B supply to the screen-grids of the two r.f. tubes. This provides a smooth control within wide limits without affecting the tuning or the tone quality.

While the screen-grid tubes have an extremely low plate-to-grid capacity, thus minimizing feedback through the tubes, this advantage is nullified if feedback occurs in other parts of the receiver. So every effort has been made to isolate all circuits in which coupling might result in instability. The negative bias for the control grids of the r.f. tubes is secured by the drop across individual 10-ohm resistors in series with the negative leg of each screen-grid tube filament. Since the screen-grids of both these tubes are biased by the 100,000 ohm potentiometer, a 5,000 ohm isolating resistor is inserted in the lead to each of the screen-grids, which are in turn by-passed by means of separate 1/2 mfd. by-pass condensers. The plate circuits of these tubes are likewise

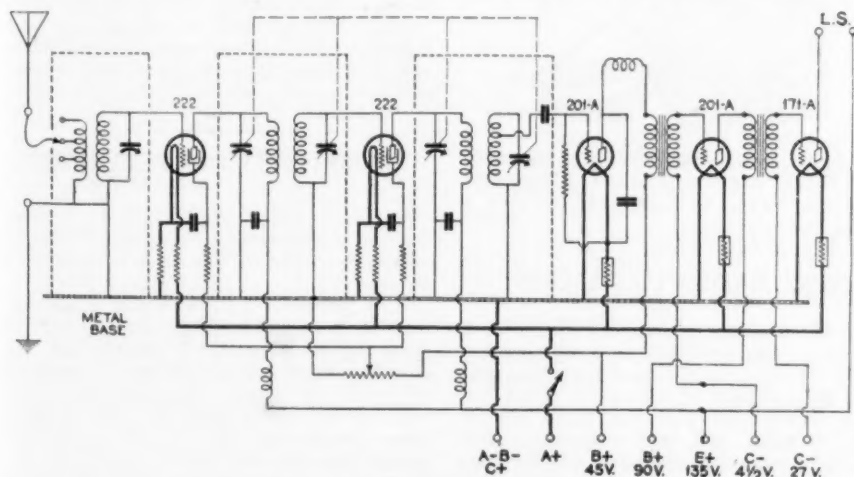
LIST OF PARTS FOR THE MASTER HI-Q 29 D. C. MODEL

- 5 Hammarlund ML-17 .00035 mfd. Midline Condensers.
- 1 Hammarlund Hi-Q 29 Coil Set.
- 2 Hammarlund SDW Knob-Control Drum Dials (Walnut).
- 3 Hammarlund RFC-85 Radio Frequency Chokes.
- 5 Benjamin Cle-Ra-Tone Sockets 9040.
- 1 Sangamo .00025 mfd. Fixed Mica Condenser.
- 1 Sangamo .001 mfd. Fixed Mica Condenser.
- 1 Electrad Type B Royalty Resistance 100,000 ohms. Potentiometer with filament switch.
- 2 Thordarson R-300 Audio Transformers.
- 4 Parvovolt .5 mfd. Series 200 Bypass Condensers.
- 1 Durham Metallized Resistor 1 1/2 megohms.
- 1 Yaxley 660 Cable Connector and Cable.
- 1 pr. Yaxley 422 Insulated Phone Tip jacks.
- 2 Amperites 1-A.
- 1 Amperite 112.
- 2 Eby Engraved Binding Posts.
- 1 "Hi-Q" 29 Foundation Unit Master (containing drilled and engraved Westinghouse Micarta panel, three complete aluminum shields, drilled steel chassis, shafts, binding post strips, Fahnestock clips, fixed resistance units, resistor mounts, brackets, clips, wire, screws, nuts, washers, and all special hardware required to complete receiver).

isolated by individual filters consisting of separate r.f. choke coils and bypass condensers. In addition, the entire r.f. end of the receiver is thoroughly shielded. Each stage is entirely enclosed in a snug-fitting aluminum box which is securely fastened to the metal chassis. The screen-grid tubes are so located that the leads to the control grids are as short as possible and farthest away from the plate leads, which are also short. By placing these tubes between the cans, the sides are used also as electrostatic tube shields, effectively preventing coupling between the tube elements and other parts of the circuit. This arrangement provides minimum coupling between output and input circuits, which is extremely important.

The audio-frequency amplifier is of the conventional type consisting of two stages of transformer-coupled amplification. The a.f. transformers have a flat frequency characteristic over the usual a.f. range. A r.f. choke coil is placed between the plate of the detector tube and the first a.f. transformer to prevent any stray r.f. voltages from getting into the a.f. amplifier. A 171-type tube is recommended for use in the last stage, although other types may be used if suitable A, B and C voltages are available.

Complete detailed directions for mounting and connecting the various parts are furnished with the kit and have also been published elsewhere. These are accompanied by large-scale drawings in which all connections are clearly indicated. The completed chassis can be installed in any one of a number of different cabinets or consoles so as to fit any style of home decoration. This is the latest and most efficient of the famous sets which have borne the Hammarlund-Roberts name. It will be found that only one station can be tuned in at one time, snapping in without the usual gradual increase in volume.



Circuit Diagram of Hammarlund-Roberts "Hi-Q 29"

Bosworth power amplifiers for radio and phonograph reproducers are made in a number of different sizes for various purposes. Model Amp. 1, which is illustrated herewith, uses a '26 tube in the first and two '71 tubes in push-pull for the last stage, with one '80 rectifier tube; 1A also has voltage taps for A, B, and C supply for a.c. sets. Model Amp. 2 and 2A, use a '26 tube in the first and a '50 tube in the last stage, with two '81 rectifier tubes. Models Amp. 3 and 4 are for use in auditoriums, with two or four speakers respectively, both being three-stage



amplifiers with a '27 tube in the first and two '27's push-pull in the second stage. The former uses two '50 tubes push-pull in the third stage with two '81 tubes and the latter four '50 tubes multiple push-pull and four '81 rectifier tubes. Model Mic, is a battery-operated two-stage outfit, for public address work.

The X-L link combines in one device a fused two-socket outlet, line voltage regulator, a control switch, and an antenna-ground connection for an a.c. receiver. The switch controls connection to the two sock-

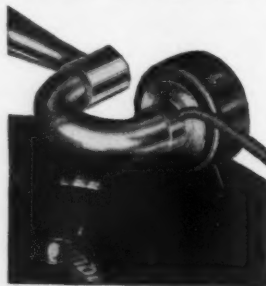


ets which may be used as the current source for the receiver and the dynamic speaker. The light wires are used as the antenna. The voltage regulator protects tubes against burn-out.

The Amperite Lin-A-trol consists of an auto-transformer and Amperite which are intended to control the line voltage supply to an a.c. receiver. The auto-transformer is designed to place a 20-volt drop across the Amperite when the line supply is 95 volts. As the line voltage increases to a normal 110 or an abnormal 125 volts, the voltage drop correspondingly increases to as high as 40 volts. But the Amperite is designed to give a nearly constant current output between 20 and 40 volts potential difference, since its resistance rapidly increases as the current output slightly increases. Depending upon the current requirements of the radio set the Amperites are made to give .4, .5, .6, .7, or .8 amperes with a variation of 10 per cent or

less, which is within the recommended limits of filament and plate supply. The entire device is compact and draws about 20 volts for its operation.

The Na-Ald electric pick-up for adapting a phonograph for use with an audio amplifier and radio speaker is equipped with two magnets and an adjusting feature to



give maximum volume without rattling. Its adapter fits either a.c. or d.c. sets with 4 or 5-prong tubes.

The Dongan radio interference filter is especially designed to eliminate interference from oil burners. It is made in two models mounted in steel cases equipped with conduit fittings for direct attachment to an oil burner. Model D-207 contains an 8 mfd. condenser and D-215, a 4 mfd. condenser.

The Wright-De Coster dynamic reproducers are made in a number of styles for 6-volt d.c. and 110-volt a.c. operation, with and without output transformers. The field coil of the chassis produces 90,000 magnetic lines per sq. in. of a.c. gap, having 2700 ampere-turns. It is claimed to handle the output of a push-pull 250 amplifier without distortion or cracking. The chassis is sold separately or in cabinets, one model also including a power plant with two 281 rectifying tubes, one 874 regulator tube, and one 250 power tube.

The Valley "B" power unit, Model 828, is a new product designed for sets using up to 10 tubes. It employs a Raytheon BH tube as a rectifier and is equipped with an over-size transformer, special chokes and high voltage condensers. It has four taps for positive plate voltages with a high resistance



control unit on the intermediate tap, and a 9 and 40 volt negative grid tap, all mounted under the cover. It is designed to give a maximum of 180 volts.

NEW RADIO CATALOGS

The Electric Supply Company of Oakland, California, has published a comprehensive catalog of radio sets, accessories and parts, which they supply at wholesale prices for resale. The parts and kits listed are of standard make and well advertised and comprise everything necessary for the construction and servicing of a.c. and d.c. receivers.

BOOK REVIEWS

Radio Library, 5 volumes, 5 x 7½ in., about 120 pp. in each volume; each subject covered by a specialist and all edited by J. F. Witkowski and F. H. Doane; published by International Correspondence Schools, Scranton, Pa. Price \$7.50.

THIS compact and convenient compilation of radio information is intended for home study by those who have the equivalent of a grammar school education. In general, the subjects are clearly and logically presented and the entire treatment is sufficient to give a good working knowledge to the practical man.

Vol. I is in two parts: "Elements of Radio Communication," by H. F. Dart and "Sources of Electromotive Force," by A. G. Zimmermann and C. H. Vose. These lay the foundation for the other volumes.

In Vol. II H. F. Dart discusses the simple principles of vacuum tubes in a conventional manner and H. H. Beverage describes the various types and properties of antennas, both transmitting and receiving. Vol. III consists of a discussion of spark, arc and tube transmitters by K. M. MacIvaine and an explanation of carrier current systems for code and sound transmissions by W. H. Freedman.

Vol. IV is devoted to a description of various standard radio receiver circuits by K. M. MacIvaine and to hints on radio servicing by the I. C. S. staff. Vol. V is in three sections: "Radio Measurements and Tables" by H. F. Dart, "Radio Developments" by E. H. Hansen and "Fire Underwriter Regulations." The second section includes information on picture transmission and television.

Each volume is well-indexed. While there is nothing novel or noteworthy in the subject matter or manner of presentation, it constitutes a comprehensive review of the past and present stages of the radio art. The material has been carefully and judiciously compiled and might well form the basis for practical application of the principles developed.

"Encyclopédie de la Radio," compiled by Michel Adam, 355 pages, 8½ x 10¾ in. Published by Étienne Chiron, 40, rue de Seine, Paris. Price, 50 francs.

This is an illustrated encyclopedia of radio terms, with definitions and explanations in French, and with English and German translations of the terms themselves. 1310 subjects are discussed rather thoroughly, in a semi-technical language that the French speaking amateur set builder can easily understand; and the 1480 illustrations showing the mechanical features and circuit drawings of various pieces of equipment make it a very helpful volume. The terms are alphabetically arranged and the explanations delve into the underlying fundamental theory wherever necessary to give the reader a more comprehensive understanding of the subject.

PUBLICATIONS RECEIVED

Research Paper No. 35, United States Bureau of Standards, "Unidirectional Radio-beacon for Aircraft," by E. Z. Stowell, price 10 cents.

This pamphlet illustrates and describes a method for transmitting signals for the guidance of aircraft. Directive and non-directive fields are simultaneously transmitted with the proper phase and amplitude relations to secure unidirectional transmission. Consequently there is only one radiated course, thus reducing interference from other beacons and increasing the power efficiency of the beacon.

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FURNISHED ROOMS

(Continued from Page 14)

the other things without affecting the
result much.

Formerly the average person was
figured as equal to 4.7 sq. ft. of open
window. But recent work has shown
that the value is 4.2. An amount equal
to 3 sq. ft. of cloth curtain has been lost
from the costume of the modern woman.
One sound engineer recently predicted
that the value will reach a new low level
of 2.5 in the near future.

Resonance in a room allows one cer-
tain note to persist long after others
have died out. Often this is caused by
small alcoves, a picture, window pane,
or vase, which may have a vibration
period within the audible range. A room
would have to be as small as the average
bathroom to have a period (frequency)
high enough to be heard. If you sing at
your bath, and have a very bad cold, you
may hit the fundamental frequency, but
otherwise not. The space between walls
may also take a notion to vibrate, as
this space is small enough to be in the
audible range.

Echo is never found in small rooms
and so will not be discussed here. Sound
must travel through two paths to get
an echo, and the difference in lengths of
path must be 65 ft. or more in order to
make one sound 1/17 second after the
other. Less than this, the ear detects
no echo.

It should be borne in mind that if
a room is acoustically corrected, it is not
soundproof. That is another story. If a
room soaks up much sound in the walls,
it may be that it is delivering this sound
to an adjoining room. Or to take the
opposite, if the common wall of two
rooms reflected 100 per cent of the sound,
the acoustics in each room would be
terrible, but each room would be sound-
proof as far as the other was concerned,
for no sound would get through the
walls. It couldn't, being all reflected
at the surface. Sound travels with a
greater speed through metals and some
solids than in air. One can see the diffi-
culty of soundproofing a house without
making the foundations and supports of
rubber or soft material.

The biggest job of soundproofing that
has been done lately is in the building
of sound stages for the talkies. Using
makeshift buildings in preliminary work,
they have had some queer results. One
photophone engineer tells the following
incident, which is a bit puzzling.

Time: 12 noon. While the cameras
grind the noon whistle blows.

The next day, time 12:20. The en-
gineer is sitting in the projection room
watching and hearing the rushes of the
previous day. Suddenly, while a love
scene is being whispered from the screen,
the noon whistle is heard. Taking out
his watch, the engineer notes that it is

(Continued on Page 34)

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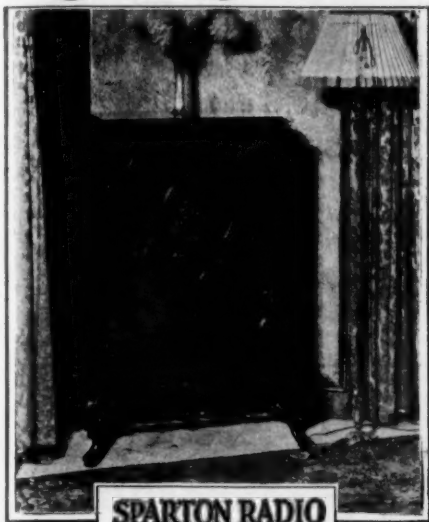
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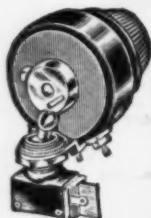


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(Continued from Page 32)

twenty minutes fast, and turns it back twenty minutes.

The next episode is sad. His train to New York leaves at four o'clock. He arrives at five minutes of four by his watch and finds that the train has been gone fifteen minutes.

We cannot see what is puzzling about this, but nine out of ten can't see what this is all about. Can you?

One curiosity in sound occurs where the room is of such a shape that it produces whispering galleries and spots where every word can be heard at another definite spot, even though there is a screen between. The capitol at Washington has the best known example. The effect is heightened because whispers are high pitched, and reflect from the walls more regularly than low pitched sounds.

If a source of high pitched sound is placed at the focus of a headlight, instead of a globe, the sound will follow the same path that the light follows (with the lens out). Whatever is whispered in front of one of two large sound reflectors at the University of California at Los Angeles can be heard only by the fair co-ed at the other reflector, an ideal arrangement. The barreled ceiling is the only curved surface commonly found in a home. If the floor is the center of the barrel the effect may be very bad.

Anyone should, then, before building a home, make sure that the architect knows a bit about what happens to sound in rooms.

SCREEN-GRID R.F. AMPLIFIER

(Continued from Page 22)

on whether the plate and grid phase angles were right for oscillations to occur. As is well known, the phase angles vary rapidly near resonance so that they might be right for oscillations. It is better in practice to keep the amplification slightly under the critical value and shield the input and output circuits from each other in order to keep the stray capacity as small as is possible.

30 K.C. TRANSFORMERS

(Continued from Page 20)

of each transformer and tied in place so that there is no strain on the No. 36 wire. Thin sheet copper may be used for making the electro-static shield. Cut two round pieces for each transformer, about 1/8 in. larger than the fiber disks. Then cut a piece 2 in. wide and long enough to go completely around the transformer with 1/4 in. overlap. Bend this piece in a circle and solder to one of the end pieces in the form of a can. Now punch four holes (for the lead wires) in the other end pieces. Place a transformer in each "can." Bring the leads through the punched holes, marking them P and S and 1 and 3 for the

(Continued on Page 36)

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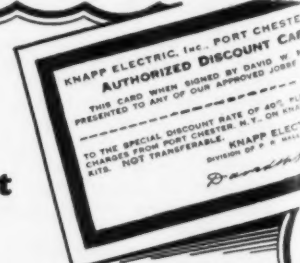
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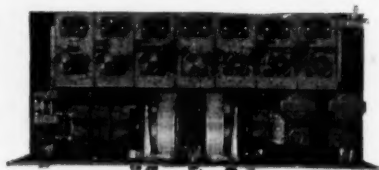
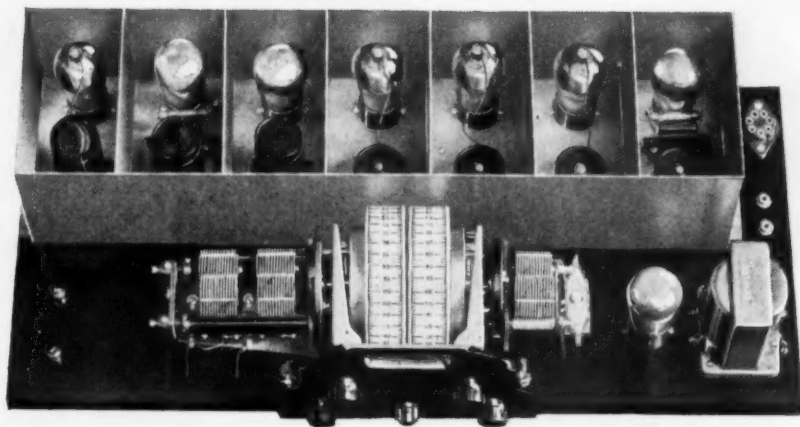
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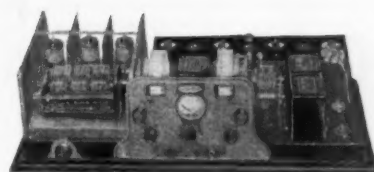
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The REMLER 29

incorporates a stage of shield-grid radio frequency amplification, oscillator, first and second detectors in the first of which regeneration is used, three stages of transformer-coupled shield-grid intermediate amplification functioning at 115 Kc., and an audio amplifier. It is to be built up on a pressed steel chassis which is drilled for Remler Audio Transformers. Remler Audio Transformers provide reproduction far superior to that which can be had from any other units. Either one or two stages of audio amplification can be built into the receiver proper. It is particularly recommended for use in cities where the usual broadcast congestion exists.



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(Continued from Page 34)

start of the windings and 2 and 4 for the finish. Now solder the end pieces to the cans in the same manner that the first ones were soldered.

When connecting in circuit, the start of the primary, *P1*, goes to the *B*-positive, the finish, *P2*, goes to the plate. The grid is connected to finish of the secondary, *S4*, and the filament to *S3*.

ANTENNA COUPLER

(Continued from Page 22)

denser—coil combination system, Fig. 4—is obtained at resonance, therefore the voltage across the combination is maximum in value, and is capacitively fed to the receiver system through the condenser *C*.

There is another advantage in the use of this circuit in preventing resonance "drags" taking place due to harmonic relation of the antenna to the oscillating detector circuit. Another advantage takes place simultaneously when the antenna is in tune with the detector circuit which is a reduction of the antenna resistance due to the energy being fed into the antenna system.

If the antenna coupling scheme is used ahead of a radio frequency amplifier, increased selectivity will be obtained with a moderate gain in signal strength due to little or no change taking place in the effective resistance of the antenna system. However, a gain is made possible by operating on the proper distribution point of the current and voltage curves.

When the coupling unit is tuned to resonance the coupling condenser *C* or

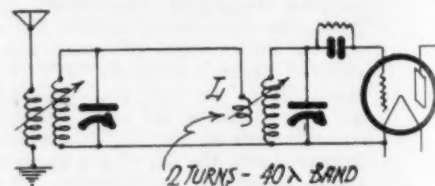


Fig. 5. More Selective Antenna-Coupling Circuit

the inductance *L*, Fig. 5, should be reduced until the detector circuit will oscillate with an increase of several divisions of the regenerative control dial. In other words, the coupling unit should not stop oscillations when both circuits are in resonance with one another. The switch *S* may be opened or closed as desired, enabling the radio fan to select a coupling combination for maximum results.

Either circuit as shown diagrammatically in Figs. 4 and 5 will bring about the desired condition as described in this article. The circuit as illustrated in Fig. 5 will enable better selectivity due to sharper tuning combinations existing in both antenna and detector circuits. It is therefore recommended for use where interference is severe. The circuit as described in Fig. 4 is easier to adjust and requires fewer coils.

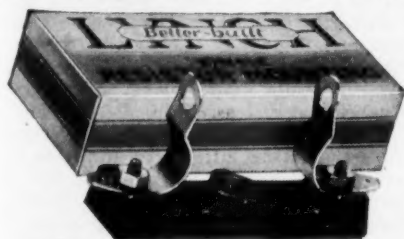
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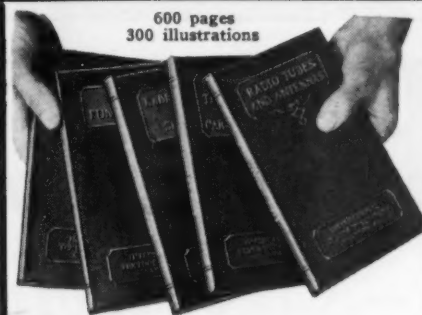
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Room 416-B, 350 Madison Ave., N. Y. C.

Tell them you saw it in RADIO

TITANIC DISASTER

(Continued from Page 25)

interrupted. Phillips advised the captain that the *Frankfurt* did not acknowledge his position, merely telling him he would see him in a few minutes.

After a lapse of about 20 minutes, during which time the *Titanic's* operator was working the *Carpathia* and *Olympic*, the *Frankfurt's* operator called again and asked what the matter was. Whether it was under the stress of strain or from being jammed, Phillips told him: "You are a fool, keep out!" and gave him the "DDD" signal, which means "QRT" in our present list of abbreviations.

The question has since been raised as to whether Phillips used good judgment in sending this, whereas, if he had sent his position and the *Frankfurt* had been close at hand it might have been possible to have saved all the lives of those on the ship. Phillips evidently thought that the *Frankfurt's* operator had received his first position report and was annoyed when he called him the second time and found that he had not. With the knowledge that the *Olympic* and *Carpathia* were steaming toward them at forced draft he perhaps felt that further talk with the *Frankfurt* would have been wasted effort, and it later turned out that he was right.

Mr. Bride then took the phones and raised the *Baltic*, while Phillips went out on deck to look around and ascertain just how badly the ship had been damaged. Bride told Phillips, when the latter returned, that the *Baltic's* signals were so weak that he did not think it worth trying further, so Phillips again took the phones while Bride went into the stateroom and proceeded to get their money together. When he came back to the operating room he saw a fireman or coalpasser trying to relieve Phillips of his lifebelt. The two of them forced him out of the cabin just as the captain came in and told them to look out for themselves, as the vessel was pretty badly damaged and would only stay afloat a few minutes longer.

They went on deck together and approached a collapsible boat which several members of the crew were trying to launch on the port side. Mr. Phillips left them, starting aft again, as he wanted to say that the engine room was being flooded, the dynamos were going out, and have a last word with the *Carpathia*. That was the last time Phillips was seen. The collapsible boat capsized as soon as it hit the water, pinning Bride and several others under it in the icy waters, where they were held for about 40 minutes before another boat finally rescued them.

Mention might be made of the fact that only four ships carried two operators at that time, the *Titanic*, *Olympic*, *Mauretania* and *Lusitania*; all the others carrying one operator each, and requiring only that watch necessary for him to handle the traffic of his vessel. It was by the merest luck that Operator Cottam of the SS. *Carpathia*, the vessel that picked up the survivors, heard the *Titanic's* call for assistance.

At 10 p. m. Cottam was receiving press from Cape Cod. The latter station finished with press at 11 p. m. and started with the *Titanic* on a long string of messages. He then took the press items that he had just copied to the bridge and spent some time there. Returning to his operating room he decided to get confirmation on a message he had sent to the coast via the SS. *Parisian* earlier in the day, providing that operator was on watch. He called several times and received no reply, so took off his coat and started to retire. Three minutes later and he should have missed the *Titanic's* call and many more lives would have been lost.

When Cottam received the CQD call of

the *Titanic* he rushed it to the bridge, returning with the *Carpathia's* position. Captain Rostron, of the *Carpathia*, figured that they were 58 miles from the *Titanic*, and at a speed of 19½ knots per hour under forced draft, would cover the distance by 4 a. m. Monday morning. During the last few minutes the *Titanic's* radio was in operation some nearby steam pipes burst, causing so much interference that it was almost impossible for the operators to work the other ships. This made it necessary for the *Carpathia* to do considerable relay work.

The *Carpathia* arrived at the scene of the disaster at 4 a. m. and took the first boatload of survivors aboard at 4:10, 20 boatloads, in all, being picked up, totaling about 720 lives. The *Carpathia* stood by during the balance of the 15th, cruising around the location in which the *Titanic* went down, but aside from a small amount of wreckage, nothing further was found. About 20 bergs, averaging from 150 to 200 feet in height, were sighted near the *Titanic's* last position, and extreme care had to be taken by the *Carpathia* in order to avoid them. The temperature at the time the *Titanic* struck was 31 degrees above zero, so it is little wonder that there was such an appalling loss from exposure. According to the final figures set by the officials of the White Star Line, 1503 people perished; many of them being of world renown. Although no positive proof has ever been submitted, it is believed that Phillips died from exposure aboard the collapsible boat. Bride was saved, but suffered severely from exposure and a sprained foot.

Cottam, operator of the *Carpathia*, remained on duty continuously from Sunday night to Tuesday night, until he was finally so completely worn out that he was unable to keep awake. Company messages were handled first, being followed by regular passenger traffic. Very little time was found for reports to the press, as the operators were kept busy with inquiry messages from ashore and reassurance messages from the passengers; so the world did not receive much information of the complete details of the disaster until the *Carpathia* tied up in New York the following Thursday.

The U. S. S. *Salem* and U. S. S. *Chester* were dispatched to the scene of the disaster by the Navy. Severe complaints and criticism were levied against the operators of the *Carpathia*, as the operators of the Naval boats claimed that the *Carpathia* discriminated and would not work them. It was shown later, however, that the third-class passenger list was sent to the operator on the U. S. S. *Chester*, but it was a long, drawn out process, as the operators on the Naval vessels used the Morse code and the *Carpathia* operators used the Continental code. This seems to be reason enough for the slight on Cottam's part.

The sinking of the *Titanic* is of special interest to the radio world because it was the primary cause of the standardization of radio procedure and other measures for safety at sea. From these measures might be mentioned the following:

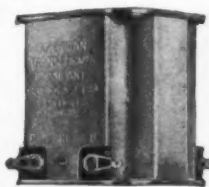
1. Adoption of the Continental Morse code as a standard for all ship operators.
2. Adoption of the conventional "Q" signals.
3. Establishment of the Ice Patrol service in the North Atlantic.
4. The requirement for a continuous watch on all passenger vessels.
5. The requirement for auxiliary means of communication and a definite range for the main set.
6. The law regarding intercommunication regardless of the system employed. (It was a well known fact that in the early days a great deal of animosity existed between operators of competitive companies.)
7. The standardization of S. O. S. as the international distress signal.

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Radio Tubes

Many other results could be traced to this terrible disaster. Never in the history of mankind has there been one single event that has caused so thorough a re-vamping of all laws governing safety at sea. Undoubtedly thousands of lives have since been safeguarded through this wise legislation.

One more item of interest that might be mentioned in closing is that at the time of the *Titanic* disaster there were only four radio equipped American ships, each carrying one operator at a salary of \$45.00 per month. These were the SS. *St. Paul*, SS. *St. Louis*, SS. *Philadelphia* and SS. *New York*.

Note from Harry Washburn:

Due to the fact that 2UO was considered out of place in the amateur band a commercial license was obtained and wavelengths of 47.13, 36.45, 27.17, and 18.226 meters were assigned. Five hundred cycle acw is still used and the transmitter is crystal controlled on all waves. Rates are 10 cents a word, with a minimum of one buck.

Another short wave commercial station is WSL, Commercial Wireless, Inc., Sayville, Long Island, N. Y. Waves: 17.805, 26.785, 35.6, 53.57 meters. Rates 10 cents, LL same as WSH.

Bill Pond, NIZJ, writes that he can't see why some of the gang insist upon jamming the weather. Sometimes we feel that there is malice aforethought, but when we cool off we realize that these operators just lack the vision to see beyond their own bulkheads. You would think that a radio operator, whose thoughts must necessarily extend to great distances, would keep in mind the doings of his neighbors. But such is not always the case.

From Charles C. Coffin, NPR:

NPR, Dutch Harbor, is situated on the Bering Sea side of the Aleutian Islands. Handles traffic from ships in the North Pacific and from the canneries in the vicinity during the fishing season. Sends local WX QST on 600 meters at 12:30 p. m. PST, and 9:30 p. m. PST. Also upon request.

Mr. Coffin wishes to pass along a word received from NPG that the GOVT weather reports from ships were reaching there too late. This can be remedied if these messages are given to either St. Paul, NPQ, or Dutch Harbor, NPR, at or as nearly after 4 a. m. and 4 p. m. PST, as possible.

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The Model 528 is also made as Ammeters which are especially useful in checking the total load of the A. C. Set—in conformity with set manufacturers' instructions. The determination of A. C. filament flow in A. C. tube filament circuits is easily obtained by means of this instrument.

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WESTON

RADIO
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Tell them you saw it in RADIO

Present Radio Conditions

Sir: I have been reading with interest Mr. Gillis' article on "Why the Present Radio Conditions," and I most certainly agree with him. I am by no means an old-timer, having been going to sea for less than two years. Mr. Gillis says he doesn't know what is causing the present conditions, but perhaps he hasn't investigated. The following are a few incidents that tend to clear matters up.

Last year I was senior on a certain coast-wise two-man job in the east. One day the senior from another company ship asked if I'd trade jobs with him because he'd had an argument with the skipper. Please bear in mind that this man is about 40 years old, and has had 17 years sea experience. I went aboard his ship with him and found out quite a few surprising things. The ship had a three-day run but he refused to copy press. He'd been on the ship for three years. Many passengers asked for baseball scores and he told them he did not have to copy press. He showed me nine letters from the Navy and Department of Commerce officials in which they told him he did not have to copy press. He told me he had written them and asked if the law required copying said press.

The junior operator was forbidden to touch the transmitter. All he was supposed to do was keep watch on 600. If a message had to be sent during the junior's watch he had to wake the senior. I refused to trade ships for fear of consequences.

I left that ship and went on a tanker. The first two weeks aboard were everything but rosy. When the crew found out I could mind my own business things got much better. The other man was a drunkard; he got press when he felt like it and considered himself superior to the other officers. As the skipper put it he was "a pain in the neck." The radio apparatus showed worse. It was dirty, below normal efficiency and clearly bespoke the man's character and habits.

Experience No. 3 was still worse. The receiver didn't work; the transformers had been removed, and some of the dollar type put in, and the shack was filthy.

These three cases happened on the three ships I've been on so I guess there are plenty such. It seems to me the service companies should fire careless operators after the second or third offense.

Yours for the B. P. section and RADIO.....
ROGER SOUCASSE.

Without a doubt such cases as those cited in Mr. Soucassee's letter do their share in retarding the brasspounding profession all over the world. The pitiful part of it is that such instances as these make a greater impression on the average outsider than instances of operators who are loyal to their profession.

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Six radio airways stations, each costing \$30,000, are to be installed by the Department of Commerce along the Pacific airway at Portland, Medford, Redding, San Francisco, Fresno, Bakersfield and Los Angeles. The purpose of the stations will be to keep commercial planes equipped with telephone apparatus in constant touch with the ground for disseminating weather and field conditions, on a wavelength of 900 meters. Commercial plane owners are advised to equip their ships so they can receive this service. These stations will not duplicate the work of private stations now operated by the Pacific Air Transport Company in connection with its operation of air mail planes, nor will it alter the work of the weather bureau in furnishing weather reports to local airports along the route.

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FREQUENCY CONTROL

(Continued from Page 26)

The imposition of supersonic sound waves between the reflecting surfaces may be controlled by enclosing the crystal and holder in a vacuum. The capacity of the holder may be kept constant by using the metal-coated crystals. Since both of these devices are beyond the reach of the average amateur we may regard the effects of both mentioned causes as being constant under normal working conditions.

It is well known that the frequency of a piezo-oscillator increases as the mechanical load upon the crystal is increased or in other words, as we increase the weight of the movable plate of the holder or the tension of the spring retaining this plate in place we will also increase the frequency at which our crystal will work. We may control the frequency to some degree by adjusting the mechanical load on the crystal. But the oscillator must be adjusted after our crystal holder is adjusted.

COMMUNICATIONS IN THE NAVAL RESERVE

A BATTLE FLEET is no better than its communication system." This statement has been ruefully made after more than one naval disaster, and as recently as the late World War. In 1916 the United States Navy mustered ships more rapidly than it could train men to handle the radio equip-



Radio W 6 C T M

ment. The fleet was 7000 operators short, and insufficiently trained men were unable to cope with the difficult situations confronting them. Some operators never got on the air; some who were more successful with the equipment, had such a vague knowledge regarding naval procedure that they did more harm than good. So the United States Navy Department has made extensive plans in the effort to eliminate the same chaotic condition should history ever repeat itself.

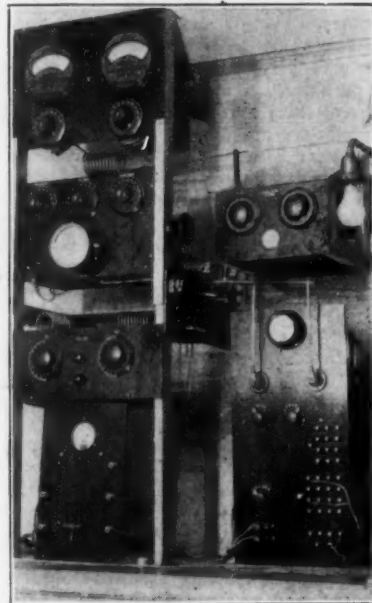
The communications unit of the United States Naval Reserve has therefore been organized to train radio operators in the use of naval operating methods and tactical procedure; to train those who desire to become radio operators in the fundamentals of radio theory and code; and, to give the reservist some instruction in seamanship, military drill and the manual of arms.

With the United States Naval Reserve communications unit at work, the Navy Department is assured of a communication system, in case any future need should arise, that will be entirely adequate to guide the maneuvers of a stupendous war-strength navy. And the communication reservist, on the other hand, will be assured a berth in the line of work for which he is most fitted. He will be of greater service to his country in time of war, and will be advanced accordingly.

Various sections of the Naval Reserve

Tell them you saw it in RADIO

throughout the country hold classes for instruction in all branches of naval communication. Tactical code drills are held on the air. Coding and decoding; semaphore and other types of visual signalling, the cor-



Radio W 6 N V

rect method of routing and the use of the naval abbreviations are studied. Some sections spend a part of their time on the drill field and on the water. All units are assigned a destroyer, upon which week-end cruises are made, and are open to all reservists who wish to take advantage of them. On these cruises practical instruction is given in the use of the ship's radio equipment and direction finders.

Any man between the ages of 18 and 35 who wishes to become a Communication Reservist, may apply to the Commandant of any naval district or to the Commanding Officer of any local section. Men enlisted in this branch of the United States Naval Reserve are not obligated to active duty, but may voluntarily apply for two weeks active training duty with full pay according to their ratings.

This service affords an excellent opportunity for thorough instruction in radio principles and practice, at no cost to the student. He has use of fine short-wave equipment which is provided by the Navy.

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Commercial, First Grade, Radioman, First Class.
Commercial, Second Grade, Radioman, Second Class.
Commercial, Cargo Grade, Seaman for Radioman.
*Experimental and Instruction Grade, Chief Radioman, Radioman First, Second and Third Class.

Amateur, First Grade, Radioman, Second Class.

Amateur, Second Grade, Radioman, Third Class.

Note 1.—The speed requirements for enlistments are as follows. Continental code only as required:

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Radioman, Second Class, send and receive at 18 words per minute.

Radioman, Third Class, send and receive at 15 words per minute.

All beginners not under the above classification are enlisted as Seamen for Radioman.



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RADIOLA 60, 62, 64

(Continued from Page 28)

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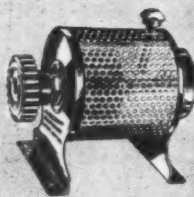
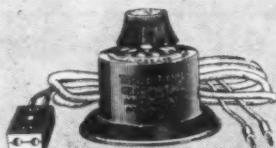
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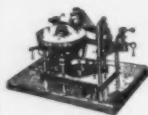
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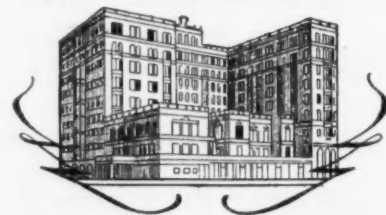
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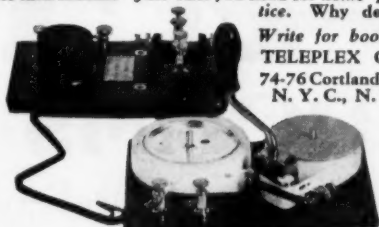
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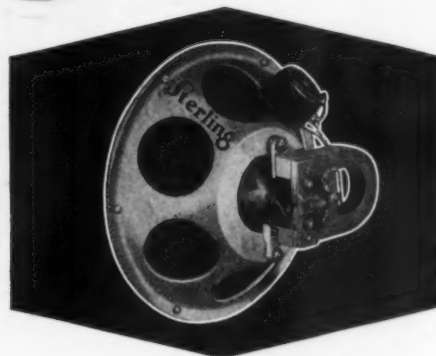
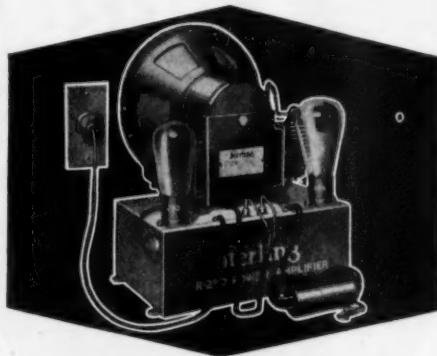
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